

# **MIC7400**

# Configurable PMIC, Five Channel Buck Regulator Plus One Boost with HyperLight Load<sup>®</sup> and I<sup>2</sup>C Control

# Features

- Input Voltage: 2.4V to 5.5V
- Five Independent Synchronous Bucks up to 3A
- One Independent Non-Synchronous Boost 200 mA
- 200 µA Quiescent Current (All Regulators On)
- 93% Peak Buck Efficiency, 85% Typical Efficiency at 1 mA
- Dual Power Modes: Stand-by and Normal Mode
- I<sup>2</sup>C Interface up to 3.4 MHz
- I<sup>2</sup>C On-the-Fly EEPROM Programmability, Featuring:
  - Buck and Boost Output Voltage Scaling
  - Power-on-Reset Threshold and Delay
  - Power-Up Sequencing/Sequencing Delay
  - Buck and Boost Current-Limit
  - Buck and Boost Pull-Down when Disabled
  - Individual ON, OFF, and Stand-by Modes
  - Soft-Start and Global Power-Good Masking
- 23 µA Buck Typical Quiescent Current
- 70 µA Boost Typical Quiescent Current
- 1.5% Output Accuracy over Temperature/Line/Load
- 2.0 MHz Boost Switching Frequency
- 1.3 MHz Buck Operation in Continuous Mode
- Ultra-Fast Buck Transient Response
- 15 mm x 15 mm x 1.25 mm Solution Size
- Thermal Shutdown and Current-Limit Protection
- 36-Pin 4.5 mm x 4.5 mm x 0.85 mm FQFN Package (0.4 mm Pitch)
- –40°C to +125°C Junction Temperature Range

# **Applications**

- Client and Enterprise Solid State Drives (SSD)
- Consumer and In-Vehicle Infotainment Devices
- Multimedia Devices
- Portable Handheld Devices
- · Security Cameras
- Gaming Machines
- Service Provider Gateways

# **General Description**

The MIC7400 is a powerful, highly integrated, configurable, power management IC (PMIC) featuring five synchronous buck regulators, one boost regulator and high-speed  $I^2C$  interface with an internal EEPROM.

The device offers two distinct modes of operation— "stand-by mode" and "normal mode"—intended to provide an energy-optimized solution suitable for portable handheld, and infotainment applications.

In normal mode, the programmable switching converters can be configured to support a variety of features, including start-up sequencing, timing, soft-start ramp, output voltage levels, current-limit levels and output discharge for each channel.

In stand-by mode the PMIC can be configured in a low power state by either disabling an output or by changing the output voltage to another voltage level, either lower or higher than normal-mode. In general, it's assumed that the voltage in standby mode is lower than the one in normal mode. Independent exit from stand-by mode can be achieved either by I<sup>2</sup>C communication or the external STBY pin.

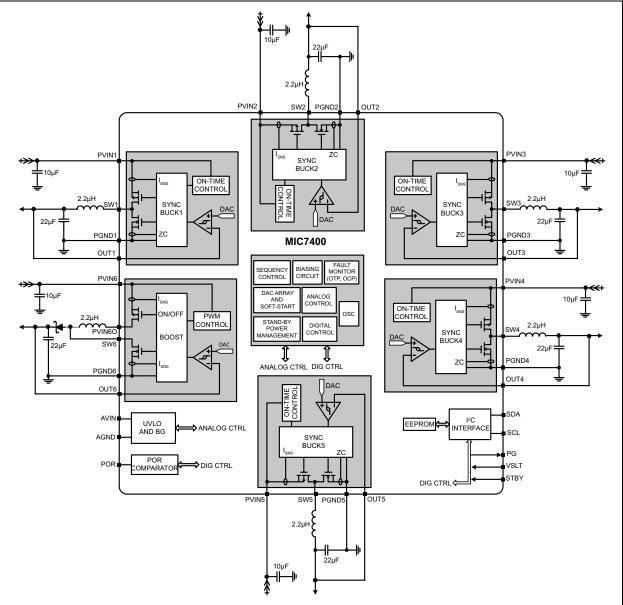
The device has five synchronous buck regulators with high-speed adaptive on-time control supporting even the challenging ultra-fast transient requirement for Core supplies. One boost regulator provides a flash memory programming supply that delivers up to 200 mA of output current. The boost is equipped with an output disconnect switch that opens if a short-to-ground fault is detected.

An internal EEPROM enables a single-chip solution across many platforms by allowing the designer to customize the PMIC for their design. Modifications can be made without the need to re-approve a new PMIC, saving valuable design resources and time.

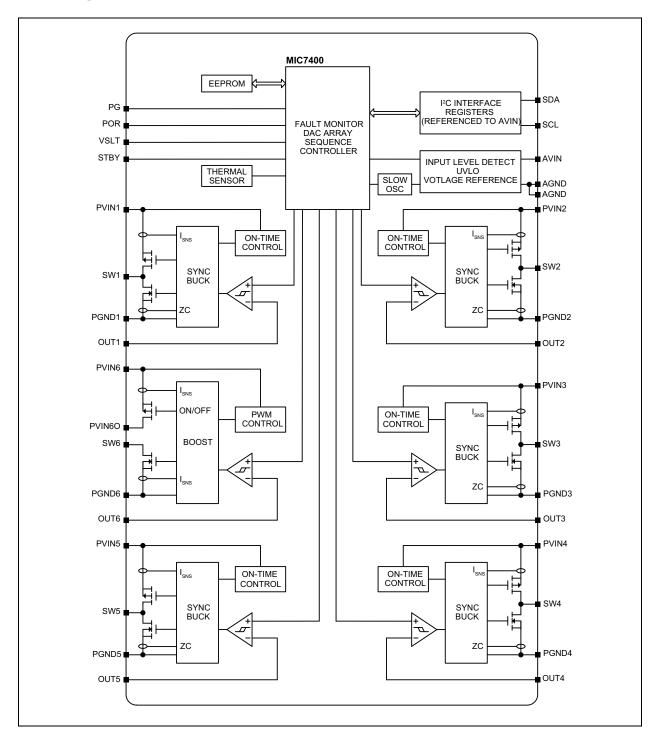
All switchers provide light load efficiency with HyperLight Load<sup>®</sup> mode for buck and PFM mode for boost. An additional benefit of this proprietary architecture is very low output ripple voltage throughout the entire load range with the use of small output capacitors. The MIC7400 is designed for use with a small inductors (down to 0.47  $\mu$ H for buck, 1.5  $\mu$ H for boost), and an output capacitor as small as 10  $\mu$ F for buck, enabling a total solution size of 15 mm x 15 mm and less than 1 mm height.

# **MIC7400**

# **Typical Application Circuit**



# **Block Diagram**



# 1.0 ELECTRICAL CHARACTERISTICS

# Absolute Maximum Ratings †

Supply Voltages (PV <sub>IN[1-6]</sub> )	–0.3V to +6V
Analog Supply Voltage (AV <sub>IN</sub> )	–0.3V to +6V
Buck Output Voltages (V <sub>OUT[1-5]</sub> )	–0.3V to +6V
Boost Output Voltage (V <sub>OUT6</sub> )	–0.3V to +20V
Buck Switch Voltages (V <sub>SW[1-5]</sub> )	–0.3V to +6V
Boost Switch Voltage (V <sub>SW6</sub> )	
Power Good Voltage (V <sub>PG</sub> )	–0.3V to AV <sub>IN</sub>
Power-On Reset Output (V <sub>POR</sub> )	–0.3V to +6V
POR Threshold Voltage (V <sub>VSLT</sub> )	–0.3V to +6V
Standby Voltage (V <sub>STBY</sub> )	–0.3V to +6V
I <sup>2</sup> C IO (V <sub>SDA</sub> , V <sub>SCL</sub> )	–0.3V to AV <sub>IN</sub>
AGND to PGND[1-6]	
ESD Rating (Note 1)	HBM: 2 kV; MM: 200V

# **Operating Ratings ‡**

Input Voltage (PV <sub>IN[1-6]</sub> )	+2.4V to +5.5V
Analog Input Voltage (AVIN)	
Buck Output Voltage Range (V <sub>OUT[1-5]</sub> )	+0.8V to +3.3V
Boost Output Voltage Range (V <sub>OUT6</sub> )	
Power Good Voltage (V <sub>PG</sub> )	0V to AV <sub>IN</sub>
Power-On Reset Output (V <sub>POR</sub> )	0V to AV <sub>IN</sub>
POR Threshold Voltage (V <sub>VSLT</sub> )	0V to AV <sub>IN</sub>
Standby Voltage (V <sub>STBY</sub> )	0V to AV <sub>IN</sub>
I <sup>2</sup> C IO (V <sub>SDA</sub> , V <sub>SCL</sub> )	

**† Notice:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability. Specifications are for packaged product only.

**‡ Notice:** The device is not guaranteed to function outside its operating ratings.

Note 1: Devices are ESD sensitive. Handling precautions are recommended. Human body model,  $1.5 \text{ k}\Omega$  in series with 100 pF.

# TABLE 1-1: ELECTRICAL CHARACTERISTICS

**Electrical Characteristics:**  $V_{IN} = AV_{IN} = PV_{IN(1-6)} = 5.0V$ ;  $V_{OUT1} = 1.8V$ ;  $V_{OUT2} = 1.1V$ ;  $V_{OUT3} = 1.8V$ ;  $V_{OUT4} = 1.05V$ ;  $V_{OUT5} = 1.25V$ ;  $V_{OUT6} = 12V$ .  $T_A = +25^{\circ}C$ , unless otherwise noted. **Bold** values indicate  $-40^{\circ}C \le T_J \le +125^{\circ}C$ . Note 1

Parameter	Min.	Тур.	Max.	Units	Conditions	
Input Supply (V <sub>IN</sub> )					·	
Input Voltage Range (AV <sub>IN</sub> , PV <sub>IN[1-6]</sub> )	2.4		5.5	V	—	
Operating Quiescent Current into AV <sub>IN</sub> (Note 2, Note 3)	_	200	240	μA	V <sub>IN</sub> = 5.0V; I <sub>OUT</sub> = 0A	
Operating Quiescent Current into PV <sub>IN</sub> (Note 2)	_	0.3	1.0	μA	V <sub>IN</sub> = 5.0V; I <sub>OUT</sub> = 0A	
Undervoltage Lockout Threshold	2.15	2.25	2.35	V	AV <sub>IN</sub> Rising	
Undervoltage Lockout Hysteresis	_	150	—	mV	—	
Standby Input (STBY)						
Logic Level High	1.2	—	—	V	—	
Logic Level Low	—	_	0.4	V	—	
Bias Current into Pin	—	—	200	nA	V <sub>STBY</sub> = V <sub>IN</sub>	
Bias Current out of Pin	_	_	200	nA	V <sub>STBY</sub> = 0V	
Rising/Falling Edge Reset Deglitch	_	100	_	μs	—	
POR Threshold Input (V <sub>SLT</sub> )						
Logic Level High	1.2				—	
Logic Level Low	—	_	0.4		—	
Bias Current Into Pin	—	—	200		$V_{VSLT} = V_{IN}$	
Bias Current Out of Pin	—	—	200		V <sub>VSLT</sub> = 0V	
Power-On-Reset (POR) Comparator						
POR Upper Comparator Range	2.646	2.7	2.754	V	AV <sub>IN</sub> Rising, V <sub>VSLT</sub> = 0V	
POR Lower Comparator Range	2.548	2.6	2.652	V	AV <sub>IN</sub> Falling, V <sub>VSLT</sub> = 0V	
POR Upper Comparator Range	3.626	3.7	3.774	V	$AV_{IN}$ Rising, $V_{VSLT} = V_{IN}$	
POR Lower Comparator Range	3.528	3.6	3.672	V	$AV_{IN}$ Falling, $V_{VSLT}$ = $V_{IN}$	
Power Reset Output (POR) and Timer						
POR Delay	18	20	22	ms	—	
POR Deglitch Delay	—	50	—	μs	AV <sub>IN</sub> Falling	
POR Output Low Voltage	_	75	400	mV	I <sub>POR</sub> = 10 mA (sinking)	
POR Leakage Current	—	—	200	nA	V <sub>POR</sub> = 5.5V	
Global Power Good Output (PG)						
Buck Power Good Threshold Voltage	87	91	95	%V <sub>OUT</sub>	V <sub>OUT[1-5]</sub> Rising	
Buck Hysteresis (Note 4)	—	4	_	%V <sub>OUT</sub>	V <sub>OUT[1-5]</sub> Falling	
Boost Power Good Threshold Voltage	87	91	95	%V <sub>OUT</sub>	V <sub>OUT[6]</sub> Rising	
Boost Hysteresis (Note 4)	—	380	—	mV	V <sub>OUT[6]</sub> Falling	
Power Good Output Low Voltage	—	75	400	mV	I <sub>PG</sub> = 10 mA (sinking)	
Power Good Leakage Current		0.01	200	nA	V <sub>PG</sub> = 5.5V	
Power Good Deglitch Delay	—	100	—	μs	V <sub>OUT[1-6]</sub> Falling	
Output Sequencing Delay (Note 4)	0.96	1	1.04	ms		
Thermal Protection	•	•	•	•	·	
Thermal Shutdown	_	160	—	°C	T <sub>J</sub> Rising	
Thermal Hysteresis		20	_	°C	—	

# TABLE 1-1: ELECTRICAL CHARACTERISTICS (CONTINUED)

**Electrical Characteristics:**  $V_{IN} = AV_{IN} = PV_{IN(1-6)} = 5.0V$ ;  $V_{OUT1} = 1.8V$ ;  $V_{OUT2} = 1.1V$ ;  $V_{OUT3} = 1.8V$ ;  $V_{OUT4} = 1.05V$ ;  $V_{OUT5} = 1.25V$ ;  $V_{OUT6} = 12V$ .  $T_A = +25^{\circ}C$ , unless otherwise noted. **Bold** values indicate  $-40^{\circ}C \le T_J \le +125^{\circ}C$ . Note 1

Parameter	Min.	Тур.	Max.	Units	Conditions		
Synchronous Buck (V <sub>OUT1</sub> - V <sub>OUT5</sub> )					·		
Buck Output Voltage Accuracy (OUT[1-5])							
Typical Output Voltage 1 Accuracy (Note 5)	-1.5	—	1.5	%	Includes Load, Line, and Reference		
Typical Output Voltage 2 Accuracy (Note 5)	-1.5	—	1.5	%	Includes Load, Line, and Reference		
Typical Output Voltage 3 Accuracy (Note 5)	-1.5	_	1.5	%	Includes Load, Line, and Reference		
Typical Output Voltage 4 Accuracy (Note 5)	-1.5	_	1.5	%	Includes Load, Line, and Reference		
Typical Output Voltage 5 Accuracy (Note 5)	-1.5	_	1.5	%	Includes Load, Line, and Reference		
Output Voltage 1 Accuracy (Note 5)	-1		1	%	—		
Output Voltage 2 Accuracy (Note 5)	-1		1	%	—		
Output Voltage 3 Accuracy (Note 5)	-1		1	%	—		
Output Voltage 4 Accuracy (Note 5)	-1		1	%	—		
Output Voltage 5 Accuracy (Note 5)	-1		1	%	—		
Load Regulation	_	0.1		%	$I_{OUT}$ = 10 mA to $I_{OUT(MAX)}$		
Line Regulation	—	0.05		%	V <sub>IN</sub> = 3.3V to 5.0V		
Buck Soft-Start					·		
Soft-Start (1-5) LSB (Note 4, Note 6)	3.84	4.0	4.16	µs/step	—		
Buck Internal MOSFETs					•		
High-Side On-Resistance	_	54		mΩ	V <sub>IN</sub> = 3.3V; I <sub>SW[1-5]</sub> = 200 mA		
High-Side On-Resistance	—	40		mΩ	V <sub>IN</sub> = 5.0V; I <sub>SW[1-5]</sub> = 200 mA		
Low-Side On-Resistance	—	37		mΩ	V <sub>IN</sub> = 3.3V; I <sub>SW[1-5]</sub> = -200 mA		
Low-Side On-Resistance	_	30		mΩ	V <sub>IN</sub> = 5.0V; I <sub>SW[1-5]</sub> = -200 mA		
Output Pull-Down Resistance	75	90	200	Ω	V <sub>SW[1-5]</sub> = 0V		
Buck Controller Timing					· · · ·		
Fixed On-Time (Note 7)	_	220		ns	V <sub>IN</sub> = 3.3; V <sub>OUT</sub> = 1.0V; I <sub>OUT</sub> = 1.0A		
Minimum OFF-Time		80		ns	—		
Buck Current-Limit (OUT1 - OUT5)					•		
Buck 1 Current-Limit Threshold	3.075	4.1	5.125	Α	See Table 4-3 for IPROG Settings		
Buck 2 Current-Limit Threshold	3.075	4.1	5.125	Α	See Table 4-3 for IPROG Settings		
Buck 3 Current-Limit Threshold	3.075	4.1	5.125	Α	See Table 4-3 for IPROG Settings		
Buck 4 Current-Limit Threshold	4.88	6.1	7.32	Α	See Table 4-3 for IPROG Settings		
Buck 5 Current-Limit Threshold	3.075	4.1	5.125	Α	See Table 4-3 for IPROG Settings		
Gross High-Side Current-Limit [1-5]	_	150	_	%	With Respect to Buck [x] Current-Limit		
Zero Cross Threshold	—	0		mV	Zero crossing detector		

# TABLE 1-1: ELECTRICAL CHARACTERISTICS (CONTINUED)

**Electrical Characteristics:**  $V_{IN} = AV_{IN} = PV_{IN(1-6)} = 5.0V$ ;  $V_{OUT1} = 1.8V$ ;  $V_{OUT2} = 1.1V$ ;  $V_{OUT3} = 1.8V$ ;  $V_{OUT4} = 1.05V$ ;  $V_{OUT5} = 1.25V$ ;  $V_{OUT6} = 12V$ .  $T_A = +25^{\circ}C$ , unless otherwise noted. **Bold** values indicate  $-40^{\circ}C \le T_J \le +125^{\circ}C$ . Note 1

Parameter	Min.	Тур.	Max.	Units	Conditions
Boost (V <sub>OUT6</sub> )					·
Boost Output Voltage (V <sub>OUT6</sub> )					
Typical Output Voltage Accuracy (Note 5)	-1.5	_	1.5	%	Includes Load, Line, and Reference
Output Voltage Accuracy (Note 5)	-1		1	%	—
Load Regulation		0.2		%	I <sub>OUT6</sub> = 1.0 mA to 200 mA
Line Regulation	_	0.2		%	V <sub>IN</sub> = 2.4V to 5.5V; I <sub>OUT6</sub> = 10 mA
V <sub>OUT6</sub> Discharge Current	111	148	185	mA	V <sub>IN</sub> = 3.3V; V <sub>OUT6</sub> = 12V
Boost Soft-Start Step Duration	•				•
Soft-Start 6 LSB (Note 4, Note 6)	3.84	4.0	4.16	µs/step	—
Boost Internal MOSFETs					
Low-Side On-Resistance		160		mΩ	V <sub>IN</sub> = 3.3V; I <sub>SW1</sub> = -100 mA
Low-Side On-Resistance		140		mΩ	V <sub>IN</sub> = 5.0V; I <sub>SW1</sub> = -100 mA
Boost Disconnect MOSFETs	1				L
Disconnect Switch On-Resistance	_	90	—	mΩ	I <sub>PVIN6O</sub> = 100 mA; V <sub>IN</sub> = 3.3V
Disconnect Switch Current-Limit		5		А	_
Boost Switching Frequency	1				
Switching Frequency (PWM Mode)	1.92	2.0	2.08	MHz	—
Minimum Duty Cycle	35	40	45	%	—
Maximum Duty Cycle	80	85	90	%	—
Boost Current-Limit	1				
NMOS Current-Limit Threshold	_	2.24	—	А	—
I <sup>2</sup> C Interface		1	1		1
I <sup>2</sup> C Interface (SCL, SDA)					
Low Level Input Voltage			0.4	V	—
High Level Input Voltage	1.2			V	—
Low Level Input Current	-200	0.01	200	nA	—
High Level Input Current	-200	0.01	200	nA	—
SDA Pull-Down Resistance	_	20		Ω	—
SDA Logic 0 Output Voltage	_		0.4	V	I <sub>SDA</sub> = 3 mA
CLK, DATA Pin Capacitance	_	0.7		pF	—
I <sup>2</sup> C Interface Timing (Note 4)	•		•	•	·
	_		100	kHz	Standard Mode
SCL Clock Frequency	_		400	kHz	Fast Mode
	—	—	3.4	MHz	High Speed Mode (Note 4)

Note 1: Specifications are for packaged product only.

- **2:** Tested in a non-switching configuration.
- **3:** When all outputs are configured to the minimum programmable voltage.
- 4: Guaranteed by design.
- **5:** Not tested in a closed loop configuration.
- 6: The soft-start time is calculated using the following equation:  $t_{softstart} = [(V_{OUT\_PROGRAM} 0.15)/0.05 + 1) \times t_{RAMP}$ .
- 7: Buck frequency is calculated using the following equation  $f_{SW} = (V_{OUT}/V_{IN}) \times (1/t_{ON})$ .

# **TEMPERATURE SPECIFICATIONS (Note 1)**

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions		
Temperature Ranges								
Junction Operating Temperature Range	Τ <sub>J</sub>	-40	_	+125	°C	—		
Ambient Storage Temperature Range	Τ <sub>S</sub>	-40	_	+150	°C	—		
Package Thermal Resistance								
Thermal Resistance FQFN-36Ld	$\theta_{JA}$	_	30	_	°C/W	—		

**Note 1:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T<sub>A</sub>, T<sub>J</sub>, θ<sub>JA</sub>). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +125°C rating. Sustained junction temperatures above +125°C can impact the device reliability.

#### 2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

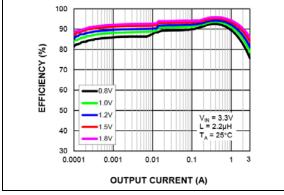


FIGURE 2-1: Buck Efficiency (LDCR = 0 mΩ) vs. Output Current.

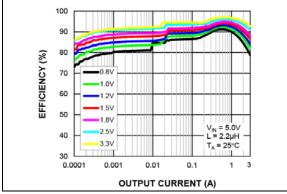
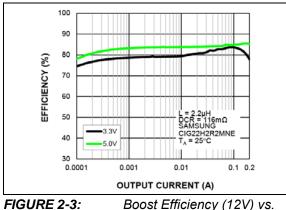
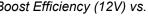


FIGURE 2-2: Buck Efficiency (LDCR = 0 mΩ) vs. Output Current.



Output Current.



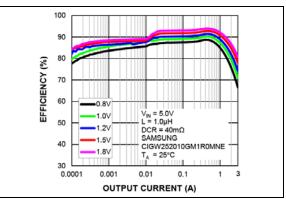


FIGURE 2-4: Buck Efficiency (LDCR = 40 mΩ) vs. Output Current.

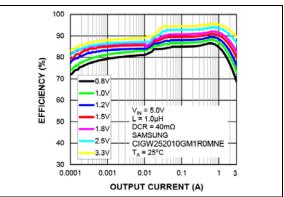
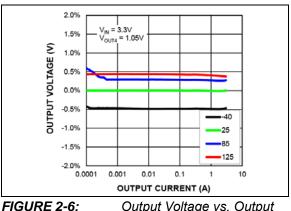
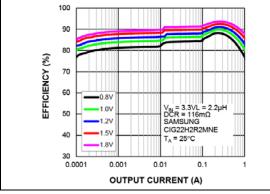


FIGURE 2-5: Buck Efficiency (LDCR = 40 mΩ) vs. Output Current.

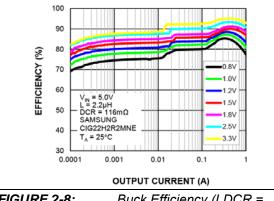


Current.

Output Voltage vs. Output



**FIGURE 2-7:** Buck Efficiency (LDCR =  $116 \text{ m}\Omega$ ) vs. Output Current.



**FIGURE 2-8:** Buck Efficiency (LDCR =  $116 \text{ m}\Omega$ ) vs. Output Current.

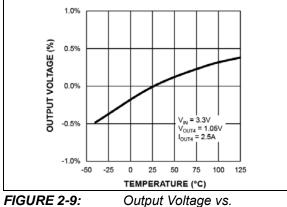
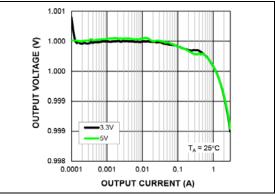


FIGURE 2-9: Temperature.



**FIGURE 2-10:** Buck Output Voltage (1.0V) vs. Output Current.

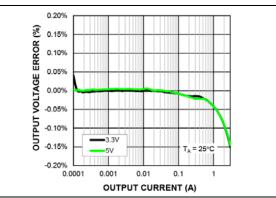


FIGURE 2-11: Buck Output Voltage Regulator vs. Output Current.

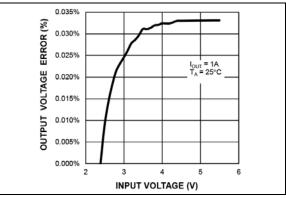
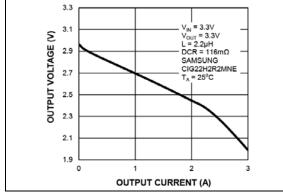
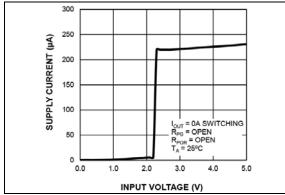


FIGURE 2-12: Buck Line Regulation vs. Input Voltage.



*FIGURE 2-13:* Dropout Output Voltage vs. Output Current.



**FIGURE 2-14:** V<sub>IN</sub> Operating Supply Current vs. Input Voltage.

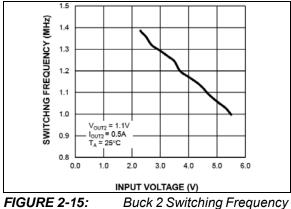
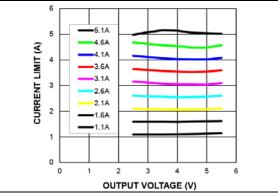


FIGURE 2-15: Buck 2 S vs. Input Voltage.



*FIGURE 2-16:* Current-Limit Threshold vs. Output Voltage.

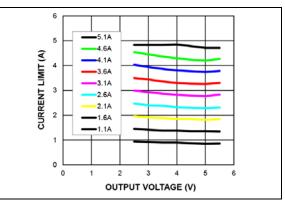
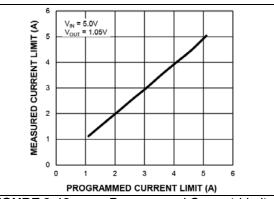
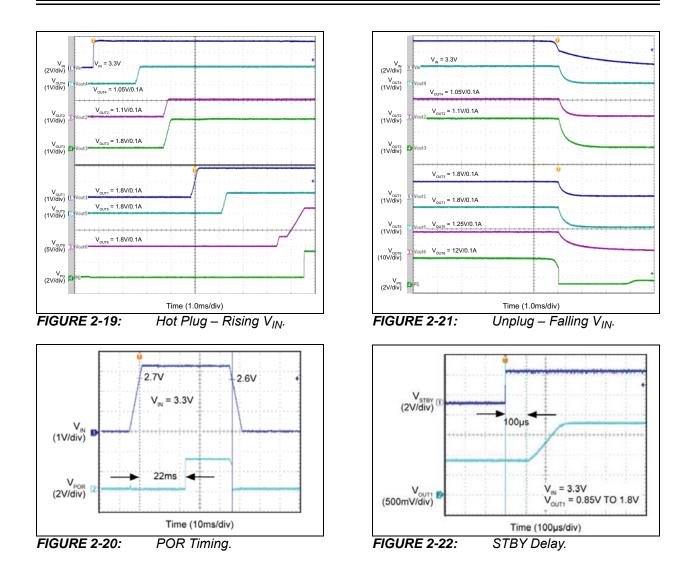
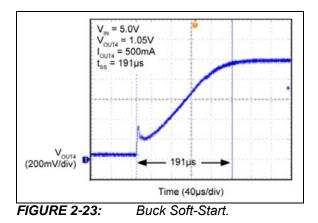


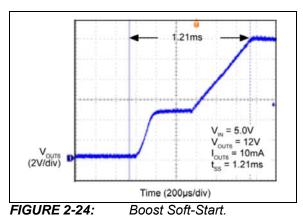
FIGURE 2-17: Output Current-Limit vs. Output Voltage.

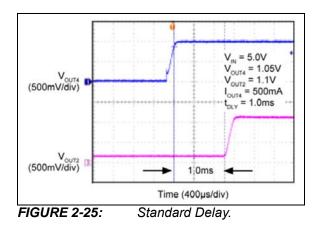


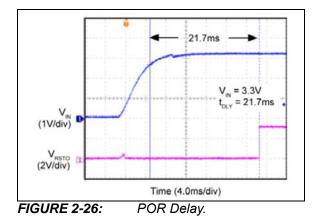
**FIGURE 2-18:** Programmed Current-Limit vs. Measured Current-Limit.

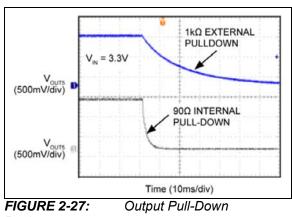




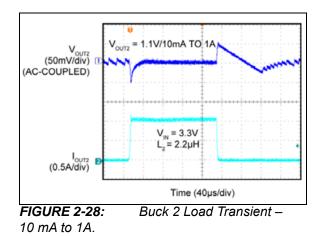












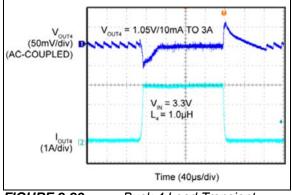
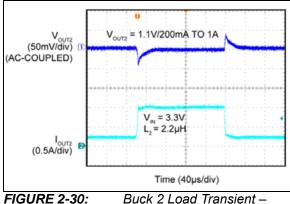
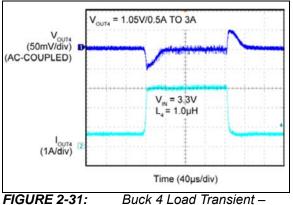


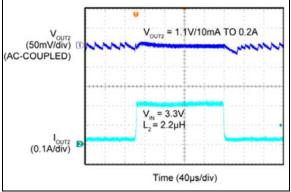
FIGURE 2-29: Buck 4 Load Transient -10 mA to 3A.



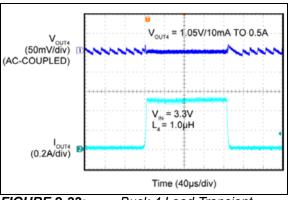
**FIGURE 2-30:** 200 mA to 1A.



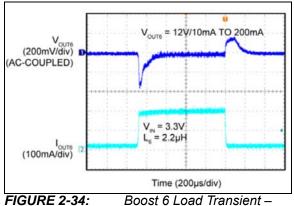
0.5A to 3A.



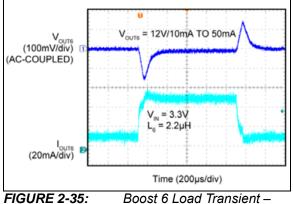
**FIGURE 2-32:** Buck 2 Load Transient -10 mA to 0.2A.



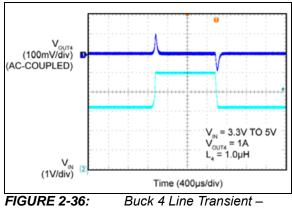
**FIGURE 2-33:** Buck 4 Load Transient -10 mA to 0.5A.



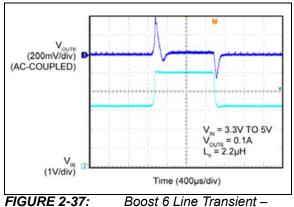
10 mA to 200 mA.



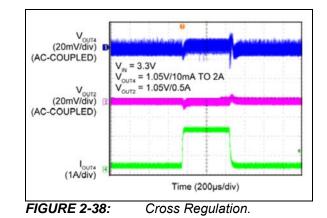
10 mA to 50 mA.



3.3V to 5.0V.



3.3V to 5.0V.



(AC-COUPLED)

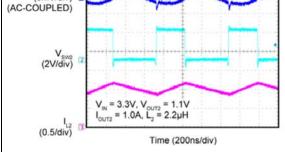


FIGURE 2-39: Buck 2 PWM Switching Waveforms.

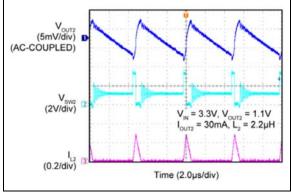


FIGURE 2-40: Waveforms.

Buck 2 PFM Switching

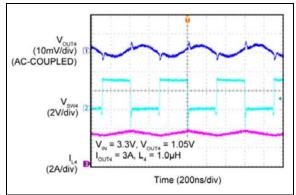


FIGURE 2-41: Buck 4 PWM Switching Waveforms.

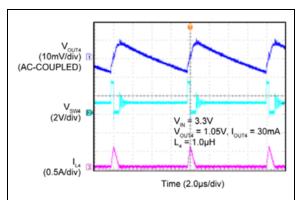


FIGURE 2-42: Waveforms.

Buck 4 PFM Switching

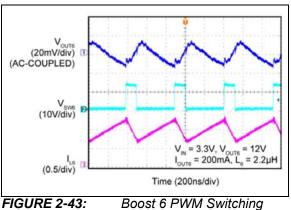


FIGURE 2-43: Waveforms.

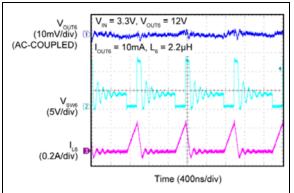


FIGURE 2-44: Boost 6 PFM Switching Waveforms.

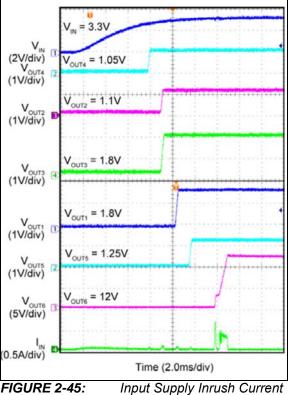
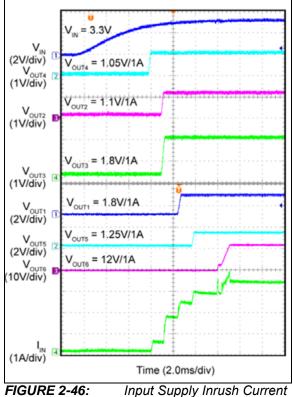
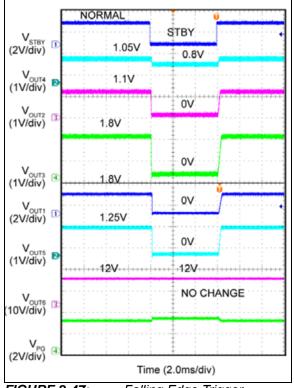


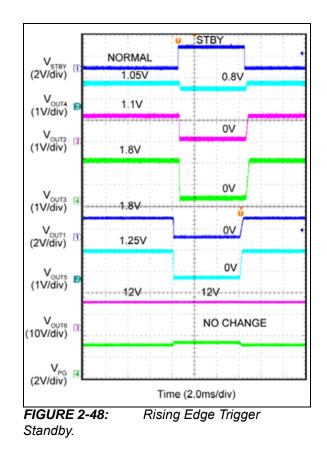
FIGURE 2-45: Input Supply Inrush – No Load.



– Loaded.



**FIGURE 2-47:** Falling Edge Trigger Standby (DEFAULT).



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# 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

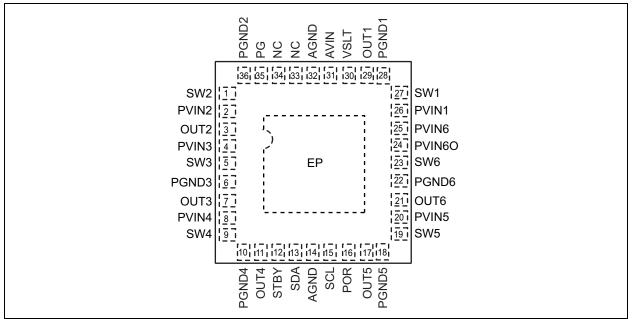


FIGURE 3-1: MIC7400 Pin Configuration.

TABLE 3-1:	PIN FUNCTION TABLE
------------	--------------------

Pin Number	Pin Name	Description
1	SW2	Switch Pin 2 (Output): Inductor connection for the synchronous step-down regulator. Connect the inductor between the output capacitor and the SW2 pin.
2	PVIN2	Power Supply Voltage 2 (Input): Input supply to the source of the internal high-side P-channel MOSFET. An input capacitor between PVIN2 and the power ground PGND2 pin is required and should be placed as close as possible to the IC.
3	OUT2	Output Voltage Sense 2 (Input): This pin is used to sense the output voltage. Connect OUT2 as close to the output capacitor as possible to sense output voltage. Also provides the path to discharge the output through an internal $90\Omega$ resistor when disabled. This pull-down feature is programmed through the PULLD[x] register.
4	PVIN3	Power Supply Voltage 3 (Input): Input supply to the source of the internal high-side P-channel MOSFET. An input capacitor between PVIN3 and the power ground PGND3 pin is required and should be placed as close as possible to the IC.
5	SW3	Switch Pin 3 (Output): Inductor connection for the synchronous step-down regulator. Connect the inductor between the output capacitor and the SW3 pin.
6	PGND3	Power Ground 3: The power ground for the synchronous buck converter power stage. The PGND pin connects to the sources of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
7	OUT3	Output Voltage Sense 3 (Input): This pin is used to sense the output voltage. Connect OUT3 as close to the output capacitor as possible to sense output voltage. Also provides the path to discharge the output through an internal $90\Omega$ resistor when disabled. This pull-down feature is programmed through the PULLD[x] register.
8	PVIN4	Power Supply Voltage 4 (Input): Input supply to the source of the internal high-side P-channel MOSFET. An input capacitor between PVIN4 and the power ground PGND4 pin is required and to be placed as close as possible to the IC.
9	SW4	Switch Pin 4 (Output): Inductor connection for the synchronous step-down regulator. Connect the inductor between the output capacitor and the SW4 pin.

Pin Number	Pin Name	Description
10	PGND4	Power Ground 4: The power ground for the synchronous buck converter power stage. The PGND pin connects to the source of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
11	OUT4	Output Voltage Sense 4 (Input): This pin is used to sense the output voltage. Connect the OUT4 as close to the output capacitor as possible to sense output voltage. Also provides the path to discharge the output through an internal $90\Omega$ resistor when disabled. This pull-down feature is programmed through the PULLD[x] register.
12	STBY	Standby Reset (Input): Standby mode allows the total power consumption to be reduced by either lowering a supply voltage or turning it off. The IC can be placed in standby mode while operating in normal mode by a high-to-low transition (DEFAULT) on the STBY input. When this occurs, the STBY_MODEB bit will be set to logic "0". Either a low-to-high transition on the STBY pin or an I <sup>2</sup> C write command to the STBY_MODEB bit sets all of the regulators to their normal mode default settings. This pin can be driven with either a digital signal or open collector output. Do not let this pin float. Connect to ground or VIN. A pull-down resistor of 100 k $\Omega$ or less can also be used. There are both a high-to-low (DEFAULT) and low-to-high normal to standby trigger options available.
13	SDA	High-Speed Mode 3.4 MHz I <sup>2</sup> C Data (Input/Output): This is an open-drain, bidirectional data pin. Data is read on the rising edge of the SCL and data is clocked out on the falling edge of the SCL. External pull-up resistors are required.
14	AGND	Analog Ground: Internal signal ground for all low power circuits. Connect to ground plane for best operation.
15	SCL	High-Speed Mode 3.4 MHz I <sup>2</sup> C Clock (Input): I <sup>2</sup> C serial clock line open-drain input. External pull-up resistors are required.
16	POR	Power-on-Reset (Output): This is an open-drain output that goes high after the POR delay time elapses. The POR delay time starts as soon as the AVIN pin voltage rises above the upper threshold set by the PORUP register. The POR output goes low without delay when AVIN falls below the lower threshold set by the PORDN register.
17	OUT5	Output Voltage Sense 5 (Input): This pin is used to sense the output voltage. Connect OUT5 as close to the output capacitor as possible to sense output voltage. Also provides the path to discharge the output through an internal $90\Omega$ resistor when disabled. This pull-down feature is programmed through the PULLD[x] register.
18	PGND5	Power Ground 5: The power ground for the synchronous buck converter power stage. The PGND pin connects to the source of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
19	SW5	Switch Pin 5 (Output): Inductor connection for the synchronous step-down regulator. Connect the inductor between the output capacitor and the SW5 pin.
20	PVIN5	Power Supply Voltage 5 (Input): Input supply to the source of the internal high-side P-channel MOSFET. An input capacitor between PVIN5 and the power ground PGND5 pin is required and should be placed as close as possible to the IC.
21	OUT6	Output Voltage 6 Sense (Input): This pin is used to sense the output voltage. Connect OUT6 as close to the output capacitor as possible to sense output voltage. Also provides the path to discharge the output through an internal programmable current source when disabled. This pull-down feature is programmed through the PULLD[x] register.
22	PGND6	Power Ground 6: The power ground for the boost converter power stage. The PGND pin connects to the source of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
23	SW6	Switch Pin 6 (Input): Inductor connection for the boost regulator. Connect the inductor between the PVIN6O and SW6 pin.

TABLE 3-1: PIN FUNCTION TABLE (CONTINUED)

# TABLE 3-1: PIN FUNCTION TABLE (CONTINUED)

Pin Number	Pin Name	Description
24	PVIN6O	Power Supply Voltage 6 (Output): This pin is the output of the power disconnect switch for the boost regulator. When the boost regulator is on, an internal switch provides a current path for the boost inductor. In shutdown, an internal P-channel MOSFET is turned off and disconnects the boost output from the input supply. This feature eliminates current draw from the input supply during shutdown. An input capacitor between PVIN6O and the power ground PGND6 pin is required and place as close as possible to the IC.
25	PVIN6	Power Supply Voltage 6 (Input): Input supply to the internal disconnect switch.
26	PVIN1	Power Supply Voltage 1 (Input): Input supply to the source of the internal high-side P-channel MOSFET. An input capacitor between PVIN1 and the power ground PGND1 pin is required and should be placed as close as possible to the IC.
27	SW1	Switch Pin 1 (Output): Inductor connection for the synchronous step-down regulator. Connect the inductor between the output capacitor and the SW1 pin.
28	PGND1	Power Ground 1: The power ground for the synchronous buck converter power stage. The PGND pin connects to the source of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
29	OUT1	Output Voltage Sense 1 (Input): This pin is used to sense the output voltage remotely. Connect OUT1 as close to output capacitor as possible to sense output voltage. This feature also provides the path to discharge the output through an internal $90\Omega$ resistor when disabled. The pull-down feature is programmed through the PULLD[x] register.
30	VSLT	POR Selection Threshold (Input): A high on this pin sets the PORUP and PORDN registers to their upper threshold limits and a low to their lower threshold limits. Do not leave floating.
31	AVIN	Analog Voltage Supply (Input): The start-up sequence begins as soon as the AVIN pin voltage rises above the IC's UVLO upper threshold. The outputs do not turn off until AVIN pin voltage falls below the lower threshold limit. A 2.2 $\mu$ F ceramic capacitor from the AVIN pin to AGND pin must be placed next to the IC.
32	AGND	Analog Ground: Internal signal ground for all low power circuits. Connect directly to the layer 2 ground plane. Layer 2 is the point where all the PGNDs and AGND are connected. Do not connect PGND and AGND together on the top layer.
33	NC	No Connect. Must be left floating.
34	NC	No Connect. Must be left floating.
35	PG	Global Power Good (Output): This is an open-drain output that is pulled high when all the regulator power good flags are high. If an output falls below the power good threshold or a thermal fault occurs, the global power good flag is pulled low. There is a falling edge de-glitch time of 50 µs to prevent false triggering on output voltage transients. A power good mask feature programmed through the PGOOD_MASK[x] registers can be used to ignore a power good fault. When masked an individual power good fault will not cause the global power good output to de-assert. Do not connect the power good pull-up resistor to a voltage higher than AVIN.
36	PGND2	Power Ground 2: The power ground for the synchronous buck converter power stage. The PGND pin connects to the source of the internal low-side N-Channel MOSFET, the negative terminals of input capacitors, and the negative terminals of output capacitors.
EP	ePAD	Exposed Pad: Must be connected to the GND plane for full output power to be realized.

# 4.0 FUNCTIONAL DESCRIPTION

The MIC7400 is one of the industry's most-advanced PMIC designed for solid state drives (SSD) on the market today. It is a multi-channel solution that offers software-configurable soft-start, sequencing, and digital voltage control (DVC) that minimizes PC board area. These features usually require a pin for programming. However, this approach makes the IC larger by increasing pin count, and also increases BOM cost due to the external components.

The following is a complete list of programmable features:

- Buck output voltage (0.8V 3.3V/50 mV steps)
- Boost output voltage (7.0V 14V/ 200 mV steps)
- Power-on-reset (2.25V 4.25V/50 mV steps)
- Power-on-reset delay (5 ms 160 ms/5 ms steps)
- Power-up sequencing (6 time slots)
- Power-up sequencing delay (0 ms 7 ms/1 ms steps)
- Soft-start (4 µs 1024 µs per step)
- Buck current limit threshold
  - (1.1A to 6.1A/0.5A steps)
- · Boost current limit threshold
- (1.76A to 2.6A/0.12A steps)
- Boost pull-down (37 mA to 148 mA/37 mA steps)
- Buck pull-down (90Ω)
- Buck standby output voltage programmable
- · Buck standby programmable standby current limit
- · Boost standby output voltage programmable
- Boost standby programmable standby current limit
- Global power good masking

These features give the system designer the flexibility to customize the MIC7400 for their application. For example,  $V_{OUT1}$  current-limit can be programmed to 4.1A and  $V_{OUT2}$  can be set to 1.1A. These outputs can be programmed to come up at the same time or 2.0 ms apart. In addition, in power-saving standby mode, the outputs can either be turned off or programmed to a lower voltage. With this programmability, the MIC7400 can be used in multiple platforms.

The MIC7400 buck regulators are adaptive on-time synchronous step-down DC-to-DC regulators. They are designed to operate over a wide input voltage range from 2.4V to 5.5V and provide a regulated output voltage at up to 3.0A of output current. An adaptive on-time control scheme is employed to obtain a constant switching frequency and to simplify the control compensation. The device includes an internal soft-start function which reduces the power supply input surge current at start-up by controlling the output voltage rise time.

The MIC7400 has a current-mode boost regulator that can deliver up to 200 mA of output current and only consumes 70  $\mu$ A of quiescent current. The 2.0 MHz switching frequency allows small chip inductors to be used. Programmable overcurrent sensing protects the boost from overloads and an output disconnect switch opens to protect against a short-circuit condition. Soft-start is also programmable and controls both the rising and falling output.

# 4.1 Programmable Buck Soft-Start Control

The MIC7400 soft-start feature forces the output voltage to rise gradually, which limits the inrush current during start-up. A slower output rise time will draw a lower input surge current. The soft-start time is based on the least significant bit (LSB) of an internal DAC and the speed of the ramp rate, as shown in Figure 4-1. This illustrates the soft-start waveform for all five synchronous buck converters. The initial step starts at 150 mV and each subsequent step is 50 mV.

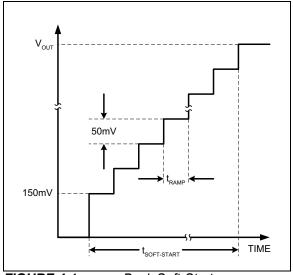


FIGURE 4-1: Bud

Buck Soft-Start.

The output ramp rate ( $t_{RAMP}$ ) is set by the soft-start registers. Each output ramp rate can be individually set from 4 µs to 1024 µs, see Table 4-1 for details.

The soft-start time  $t_{SS}$  can be calculated by Equation 4-1:

**EQUATION 4-1:** 

$$t_{SS} = \left(\frac{V_{OUT} - 0.15V}{50mV}\right) \times t_{RAMP}$$

Where:

 $t_{SS}$  = Output rise time. V<sub>OUT</sub> = Output voltage.  $t_{RAMP}$  = Output dwell time. For example:

# **EQUATION 4-2:**

$$t_{SS} = \left(\frac{1.8V - 0.15V}{50mV}\right) \times 8\mu s$$

 $t_{SS} = 264 \mu s$ 

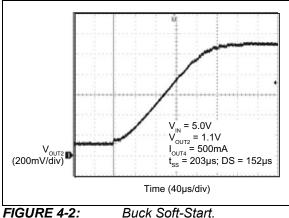
Where:

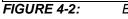
V<sub>OUT</sub> = 1.8V t<sub>RAMP</sub> = 8.0 μs

#### **TABLE 4-1: BUCK OUTPUTS DEFAULT** SOFT-START TIME (DEFAULT)

—	V <sub>OUT</sub>	t <sub>RAMP</sub>	t <sub>SS</sub>
V <sub>OUT1</sub>	1.8V	8 µs	264 µs
V <sub>OUT2</sub>	1.1V	8 µs	152 µs
V <sub>OUT3</sub>	1.8V	8 µs	264 µs
V <sub>OUT4</sub>	1.05V	8 µs	144 µs
V <sub>OUT5</sub>	1.25V	8 µs	176 µs

Figure 4-2 shows the output of Buck 1 ramping up cleanly, starting from 0.15V to its final 1.1V value.

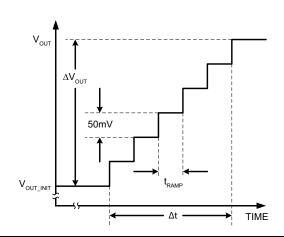




4.2

# Buck Digital Voltage Control (DVC)

The output voltage has a 6-bit control DAC that can be programmed from 0.8V to 3.3V in 50 mV increments. If the output is programmed to a higher voltage, then the output ramps up, as shown in Figure 4-3.





The ramp time is determined by Equation 4-3:

### **EQUATION 4-3:**

$$\Delta t = \left(\frac{V_{OUT} - V_{OUT\_INIT}}{50mV}\right) \times t_{RAMP}$$

Where:

V<sub>OUT\_INIT</sub> = Initial output voltage. V<sub>OUT</sub><sup>-</sup>= Final output voltage. t<sub>RAMP</sub> = Output dwell time.

When the regulator is programmed to a lower voltage, then the output voltage ramps down at a rate determined by the output ramp rate (t<sub>RAMP</sub>), the output capacitance and the external load. Small loads result in slow output voltage decay and heavy loads cause the decay to be controlled by the DAC ramp rate.

In Figure 4-4,  $V_{\mbox{OUT1}}$  is switched to stand-by mode with an  $I^2C$  command and then switched back to normal mode either by an I<sup>2</sup>C command or a low-to-high transition of the STBY pin. In this case, the rise and fall times are the same due to a 1A load on V<sub>OUT1</sub>.

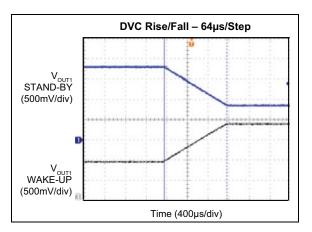
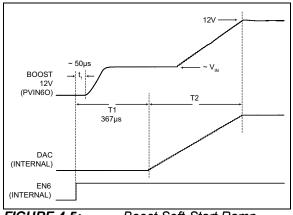


FIGURE 4-4: Buck DVC Control Ramp.

#### 4.3 **Programmable Boost Soft-Start** Control

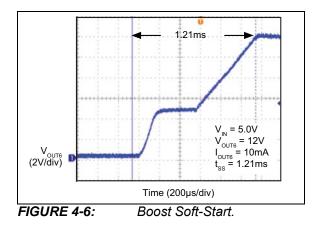
The boost soft-start time is divided into two parts as shown in Figure 4-5. T1 is a fixed 367 µs delay starting from when the internal enable goes high. This delay gives enough time for the disconnect switch to turn on and bring the inductor voltage to  $\mathsf{V}_{\mathsf{IN}}$  before the boost is turned on. There is a 50 µs delay that is controlled by the parasitic capacitance (C<sub>GD</sub>) of the disconnect switch before the output starts to rise.

After the T1 period, the DAC output ramp starts, T2. The total soft-start time, t<sub>SS</sub>, is the sum of both periods. Figure 4-6 displays the actual boost soft-start waveform.





Boost Soft-Start Ramp.



### **EQUATION 4-4:**

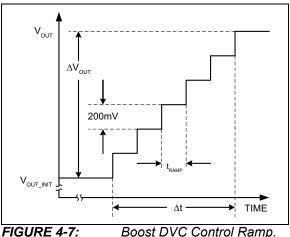
$$t_{SS} = T1 + T2$$
$$T2 = \left(\frac{V_{OUT} - 1.4V}{0.2V}\right) \times t_{RAMP}$$
$$T2 = \left(\frac{12V - 1.4V}{0.2V}\right) \times 16\mu s$$

Where:

T1 = 367 µs T2 = 848 µs  $t_{SS}$  = 367 µs + 848 µs = 1.215 ms V<sub>OUT</sub> = Output voltage.  $t_{RAMP}$  = Output dwell time = 16 µs.

#### 4.4 **Boost Digital Voltage Control** (DVC)

The boost output control works the same way as the buck, except that the voltage steps are 200 mV, see Figure 4-7. When the boost is programmed to a lower voltage the output ramps down at a rate determined by the output ramp rate (t<sub>RAMP</sub>), the output capacitance and the external load. During both the ramp up and down time, the power good output is blanked and if the power good mask bit is set to "1".



Boost DVC Control Ramp.

The ramp time can be computed using the following equation:

# **EQUATION 4-5:**

$$\Delta t = \left(\frac{V_{OUT} - V_{OUT\_INIT}}{0.2V}\right) \times t_{RAMP}$$

Where:

 $V_{OUT | INIT}$  = Initial output voltage.

# TABLE 4-2:BOOST OUTPUT DEFAULT<br/>SOFT-START TIME

_	V <sub>OUT</sub>	t <sub>RAMP</sub>	t <sub>SS</sub>
V <sub>OUT6</sub>	12V	16 µs	1.215 ms

# 4.5 Buck Current-Limit

The MIC7400 buck regulators have high-side current-limiting that can be varied by a 4-bit code. If the regulator remains in current-limit for more than seven consecutive PWM cycles, the output is latched off, the overcurrent status register bit is set to 1, the power-good status register bit is set to 0 and the global power good (PG) output pin is pulled low. An overcurrent fault on one output will not disable the remaining outputs. Table 4-3 shows the current-limit register settings verses output current. The current-limit register setting is set at twice the maximum output current.

# TABLE 4-3:BUCK CURRENT-LIMITREGISTER SETTINGS

I <sub>OUT(MAX)</sub>	I <sub>PROG</sub>	BINARY	HEX
0.5A	1.1A	1111	F'h
1.0A	2.1A	1101	D'h
1.5A	3.1A	1011	B'h
2.0A	4.1A	1001	9'h
2.5A	5.1A	0111	7'h
3.0A	6.1A	0101	5'h

The output can be turned back on by recycling the input power or by software control. To clear the overcurrent fault by software control, set the enable register bit to "0" then clear the overcurrent fault by setting the fault register bit to "0". This will clear the overcurrent and power good status registers. Now the output can be re-enabled by setting the enable register bit to "1".

During start-up sequencing, once an overcurrent condition is sensed, the fault register is set to "1" and the start-up sequence will stop and no further outputs will be enabled. See Figure 4-9 for default start-up sequence.

# 4.6 Boost Current-Limit

The boost current-limit features cycle-by-cycle protection. The duty cycle is cut immediately once the current-limit is hit. When the boost current-limit is hit for five consecutive cycles, the FAULT signal is asserted and remains asserted with the boost converter keeping on running until the boost is powered off.

This protects the boost in normal overload conditions, but not in a short-to-ground case. For a short-circuit to ground, the boost current-limit will not be able to limit the inductor current. This short-circuit condition is sensed by the current in the disconnect switch. When the disconnect switch current limit is hit for four consecutive master clock cycles (2 MHz), regardless if the boost is switching or not, both the disconnect switch and boost are latched off automatically and the FAULT signal is asserted.

The output can be turned back on by recycling the input power or by software control. To clear the overcurrent fault by software control, set the enable register bit to "0" then clear the overcurrent fault by setting the fault register bit to "0".

#### 4.7 **Global Power Good Pin**

The global power-good output indicates that all the outputs are above the 91% limit after the power-up sequence is completed. Once the power-up sequence is complete, the global power good output stays high unless an output falls below its power-good limit, a thermal fault occurs, the input voltage drops below the lower UVLO threshold or an output is turned OFF by setting the enable register bit to "0" unless the PGOOD MASK[x] bit is set to "1" (Default).

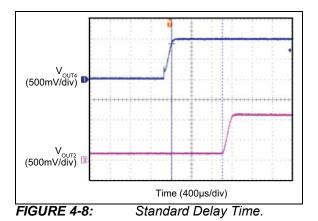
A power-good mask bit can be used to control the global power good output. The power-good mask feature is programmed through the PGOOD MASK[x] registers and is used to ignore an individual power-good fault. When masked, PGOOD MASK[x] bit is set to "1", an individual power good fault will not cause the global power good output to de-assert.

If all the PGOOD MASK[x] bits are set to "1", then the power good output de-asserts as soon as the first output starts to rise. The PGOOD MASK[x] bit of the last output must be set to "0" to have the PG output stay low until the last output reaches 91% of its final value.

The global power-good output is an open-drain output. A pull-up resistor can be connected to V<sub>IN</sub> or V<sub>OUT</sub>. Do not connect the pull-up resistor to a voltage higher than AV<sub>IN</sub>.

#### 4.8 Standard Delay

There is a programmable timer that is used to set the standard delay time between each time slot. The timer starts as soon as the previous time slot's output power good goes high. When the delay completes, the regulators assigned to that time slot are enabled, see Figure 4-8.

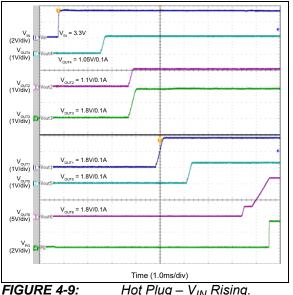


#### **Power-Up Sequencing** 4.9

When power is first applied to the MIC7400, all I<sup>2</sup>C registers are loaded with their default values from the EEPROM. There is about a 1.5 ms delay before the first regulator is enabled while the MIC7400 goes through the initialization process. The DELAY register's STDEL bits set the delay between powering up each regulator at initial power up.

The sequencing registers allow the outputs to come up in any order. There are six time slots that an output can be configured to power up in. Each time slot can be programmed for up to six regulators to be turned on at once or none at all.

Figure 4-9 shows an example of this feature. V<sub>OUT4</sub> is enabled in time slot 1. After a 1 ms delay,  $V_{OUT2}$  and V<sub>OUT3</sub> are enabled at the same time in time slot 2. The 1 ms is the standard delay for all of the outputs and can be programmed from 0 ms to 7 ms in 1 ms steps. Next,  $V_{OUT1}$  is powered up in time slot 3 and  $V_{OUT5}$  in time slot 4. There are no regulators programmed for time slot 5. Finally,  $V_{\text{OUT6}}$  is powered up in time slot 6. The global power good output,  $V_{PG}$ , goes high as soon as the last output reaches 91% of its final value.



Hot Plug – V<sub>IN</sub> Rising.

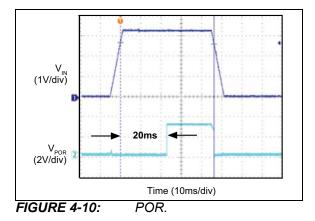
#### 4.10 VSLT Pin

The power-on-reset threshold toggles between two different ranges by driving the VSLT pin high or low. The lower range of 2.25V to 3.25V is selected when the VSLT pin is tied to ground. The upper range, 3.25V to 4.25V, is selected when the VSLT pin is tied to  $V_{IN}$ .

#### 4.11 Programmable Power-on-Reset (POR) Delay

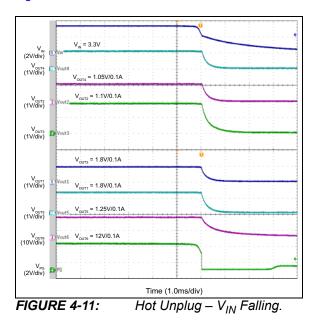
The POR output pin provides the user with a way to let the SOC know that the input power is failing. If the input voltage falls below the power-on reset lower threshold level, the POR output immediately goes low. The lower threshold is set in the PORDN register and the upper threshold uses PORUP register.

The low-to-high POR transition can be delayed from 5 ms to 160 ms in 5 ms increments. This feature can be used to signal the SOC that the power supplies are stable. The PORDEL register sets the delay of the POR pin. The POR delay starts as soon as the AVIN pin voltage rises above the power-on reset upper threshold limit. Figure 4-10 shows the POR operation.



# 4.12 Power-Down Sequencing

When power is removed from  $V_{IN}$ , all the regulators try to maintain the output voltage until the input voltage falls below the UVLO limit of 2.35V as shown in Figure 4-11.

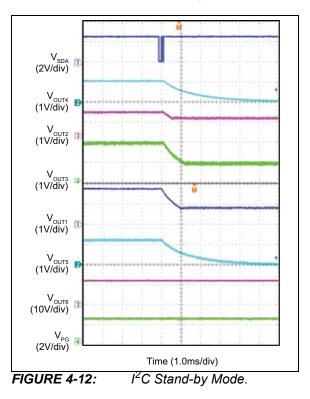


# 4.13 Stand-By Mode

In stand-by mode, efficiency can be improved by lowering the output voltage to the standby mode value or turning an output off completely. There are two registers used for setting the output voltage, normal-mode register and stand-by mode register. The default power-up voltages are set in the normal-mode registers.

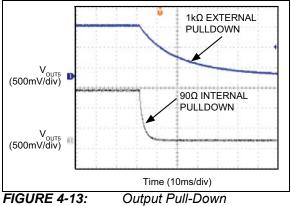
An  $I^2C$  write command to the STBY\_CTRL\_REG register or the STBY pin can be used to set the MIC7400 into stand-by mode. Figure 4-12 shows an  $I^2C$  write command implementation. In stand-by mode, the output can be programmed to a lower voltage or turned completely off. When disabled, the output will be soft-discharged to zero if the PULLD[1-6] register are set to 1. If PULLD[x] = 0 the output drifts to PGND at a rate determined by the load current and output capacitance.

In stand-by, if an output is disabled, the global power good output is not affected when the PGOOD\_MASK[x] is set to logic 1. If the PGOOD\_MASK[x] is set to logic 0, then the global power good flag is pulled low. In Figure 4-12, all the PGOOD\_MASK[x] bits are set to logic 1.



# 4.14 Resistive Discharge

To ensure a known output condition in stand-by mode, the output is actively discharged to ground if the output is disabled. Setting the buck pull down register field PULLD[1-5] = 1 connects a 90 $\Omega$  pull down resistor from OUT[x] to PGND[x] when the MIC7400 is disabled. If PULLD[x] = 0 the output drifts to PGND at a rate determined by the load current and the output capacitance value. The boost has a programmable pull-down current level from 37 mA to 148 mA. In Figure 4-13, the top trace shows the normal pull down and the bottom trace is with the 90 $\Omega$  pull-down.



Resistance.

# 4.15 STBY Pin

A pin-selectable STBY input allows the MIC7400 to be placed into standby or normal mode. In standby mode, the individual regulator can be turned on or off or the output voltage can be set to a different value. If the regulators are turned off, standby mode cuts the quiescent current by 23  $\mu$ A for each buck regulator and 70  $\mu$ A for the boost.

Figure 4-14 illustrates the STBY pin operation. A low-to-high transition on the STBY pin switches the output from standby mode to normal mode. There is a 100  $\mu$ s STBY de-glitch time to eliminate nuisance tripping then all the regulators are enabled at the same time and ramp up with their programmed ramp rates. A high-to-low transition on the STBY pin switches the output from normal mode to standby mode.

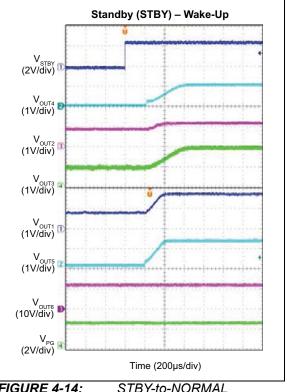
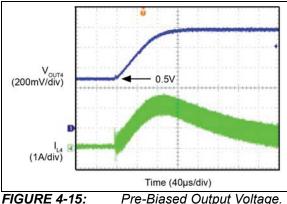


FIGURE 4-14: STBY-to-NORMAL Transition (DEFAULT).

# 4.16 Safe Start-Up into a Pre-Biased Output

The MIC7400 is designed for safe start-up into a pre-biased output. This prevents large negative inductor currents that can cause the output voltage to dip and excessive output voltage oscillations. A zero crossing comparator is used to detect a negative inductor current. If a negative inductor current is detected, the low-side synchronous MOSFET functions as a diode and is immediately turned off.

Figure 4-15 shows a 1V output pre-bias at 0.5V at start-up, see  $V_{OUT4}$  trace. The inductor current, trace  $I_{L4}$ , is not allowed to go negative by more than 0.5A before the low-side switch is turned off. This feature prevents high negative inductor current flow in a pre-bias condition which can damage the IC.



Pre-Biased Output Voltage.

#### 4.17 **Buck Regulator Power Dissipation**

The total power dissipation in a MIC7400 is a combination of the five buck regulators and the boost dissipation. The buck regulators (OUT1 to OUT5) dissipation is approximately the switcher's input power minus the switcher's output power and minus the power loss in the inductor:

### **EQUATION 4-6:**

$$P_{D\_\text{BUCK}} \approx V_{IN} \times I_{IN} - V_{OUT} \times I_{OUT} - P_{L\_\text{LOSS}}$$

While the boost power dissipation is estimated by Equation 4-7:

# **EQUATION 4-7:**

$$\begin{split} P_{D\_\text{BOOST}} &\approx V_{IN} \times I_{IN} - V_{OUT} \times I_{OUT} - P_{L\_\text{LOSS}} \\ &- V_f \times I_{OUT} \end{split}$$

Although the maximum output current for a single buck regulator can be as much as 3A, the MIC7400 will thermal limit and will not support this high output current on all outputs at the same time.

#### 4.18 **Total Power Dissipation**

The total power dissipation in the MIC7400 package is equal to the sum of the power loss of each regulator:

### **EQUATION 4-8:**

$$P_D \text{ TOTAL} \approx SUM(P_D \text{ SWITCHERS})$$

Once the total power dissipation is calculated, the IC junction temperature can be estimated using Equation 4-9:

### **EQUATION 4-9:**

$$T_{J(MAX)} \approx T_A + P_{D_{TOTAL}} \times \theta_{JA}$$

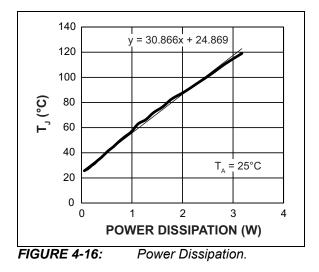
Where:

 $T_{J(MAX)}$  = The maximum junction temperature.

 $T_A = The ambient temperature.$ 

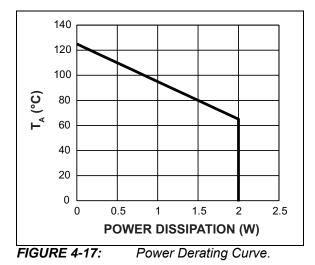
 $\theta_{JA}$  = The junction-to-ambient thermal resistance of the package (30°C/W).

Figure 4-16 shows the measured junction temperature versus power dissipation of the MIC7400 evaluation board. The actual junction temperature of the IC depends upon many factors. The significant factors influencing the die temperature rise are copper thickness in the PCB, the surface area available for convection heat transfer, air flow and power dissipation from other components, including inductors, SOCs and processor ICs. It is good engineering practice to measure all power components temperature during the final design review using a thermal couple or IR thermometer, see the Thermal Measurements sub-section for details.



# 4.19 Power Derating

The MIC7400 package has a 2W power dissipation limit. To keep the IC junction temperature below a 125°C design limit, the output power has to be limited above an ambient temperature of 65°C. Figure 4-17 shows the power dissipation derating curve.



The maximum power dissipation of the package can be calculated by Equation 4-10:

**EQUATION 4-10:** 

$$P_{D(MAX)} \approx \frac{T_{J(MAX)} - T_A}{\theta_{IA}}$$

Where:

 $T_{J(MAX)}$  = The maximum junction temperature (125°C).  $T_A$  = The ambient temperature.  $\theta_{JA}$  = The junction-to-ambient thermal resistance of the package (30°C/W).

# 4.20 Overtemperature Fault

An overtemperature fault is triggered when the IC junction temperature reaches  $160^{\circ}$ C. When this occurs, both the overtemperature fault flag is set to "1", the global power good output is pulled low and all the outputs are turned off. During the fault condition the I<sup>2</sup>C interface remains active and all registers values are maintained.

When the die temperature decreases by 20°C the overtemperature fault bit can be cleared. To clear the fault, either recycle power or write a logic "0" to the over temperature fault register. Once the fault bit is cleared, the outputs power up to their default values and are sequenced according to the time slot settings.

# 4.21 Input Voltage "Hot Plug"

High voltage spikes of twice the input voltage can appear on the MIC7400 PVIN pins if a battery pack is hot-plugged to the input supply voltage connection as shown in Figure 4-18 (Trace 1). These spikes are due to the inductance of the wires to the battery and the very low inductance and ESR of the ceramic input capacitors. This problem can be solved by placing a 150  $\mu$ F POS capacitor across the input terminals. Figure 4-18 (Trace 2) shows that the high voltage spike is greatly reduced to a value below the maximum allowable input voltage rating.

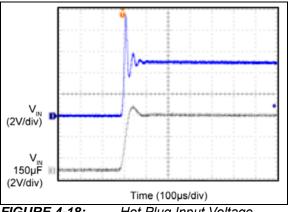


FIGURE 4-18: Hot Plug Input Voltage Spike.

# 4.22 Thermal Measurements

Measuring the IC's case temperature is recommended to ensure it is within its operating limits. Although this might seem like a very elementary task, it is easy to get erroneous results. The most common mistake is to use the standard thermal couple that comes with a thermal meter. This thermal couple wire gauge is large (typically 22 gauge) and behaves like a heatsink, resulting in a lower case measurement. Two reliable methods of temperature measurement are a smaller thermal couple wire or an infrared thermometer. If a thermal couple wire is used, it must be constructed of 36 gauge wire or higher (smaller wire size) to minimize the wire heat-sinking effect. In addition, the thermal couple tip must be covered in either thermal grease or thermal glue to make sure that the thermal couple junction is making good contact with the case of the IC. Omega brand thermal couple (5SC-TT-K-36-36) is adequate for most applications.

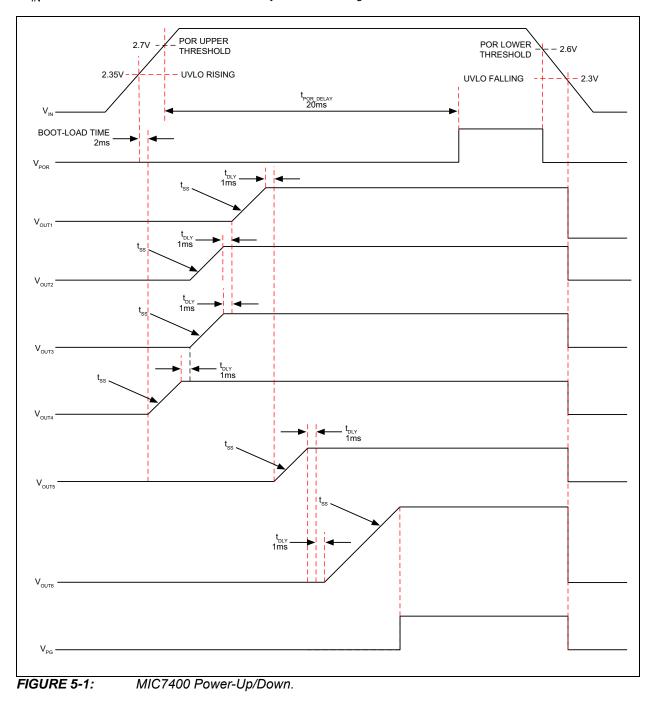
Whenever possible, an infrared thermometer is recommended. The measurement spot size of most infrared thermometers is too large for an accurate reading on a small form factor ICs. However, an IR thermometer from Optris has a 1 mm spot size, which makes it a good choice for measuring the hottest point on the case. An optional stand makes it easy to hold the beam on the IC for long periods of time.

# 5.0 TIMING DIAGRAMS

# 5.1 Normal Power-Up Sequence for Outputs

The STDEL register sets the delay between powering up of each regulator at initial power-up (see power-up sequencing in Figure 5-1). Once all the internal power good registers PGOOD[1-6] are all "1", then the global PG pin goes high without delay (see the Global Power Good Pin section for more information).

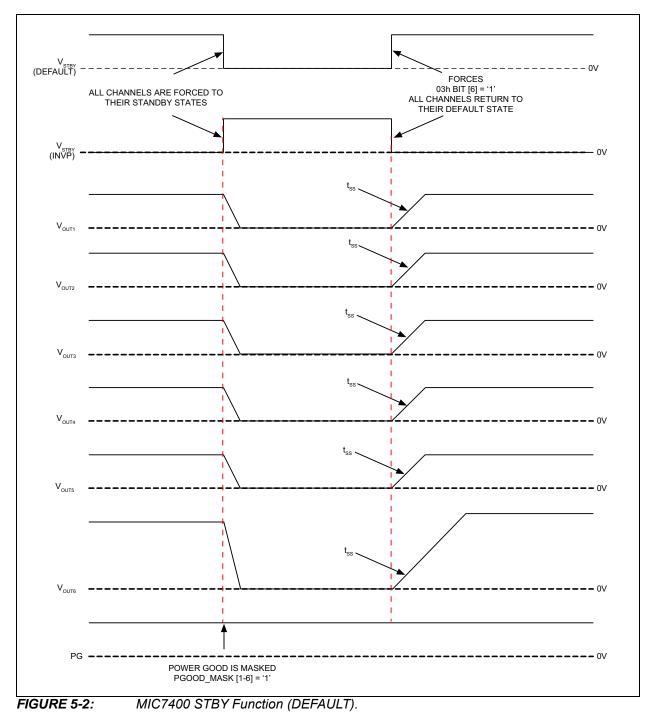
The PORDEL register sets the delay for the POR flag pin. The POR delay time starts as soon as the  $AV_{IN}$  pin voltage rises above the system UVLO upper threshold set by the PORUP register. The POR output goes low without delay if  $AV_{IN}$  falls below the lower UVLO threshold set by the PORDN register.



# 5.2 Standby (STBY) Pin (Wake-Up)

An I<sup>2</sup>C write command to the STBY\_CTRL\_REG register or the STBY pin can be used to set the MIC7400 into standby mode. The standby (STBY) pin provides a hardware-specific manner in which to wake-up from stand-by mode and go into normal mode. Figure 5-2 shows the STBY pin operation. A low-to-high transition on the STBY pin switches the output from stand-by mode to normal mode.

There is a 100  $\mu$ s STBY deglitch time to eliminate nuisance tripping, then all the regulators are enabled at the same time and ramp up with their programmed ramp rates.



# 6.0 PCB LAYOUT GUIDELINES

PCB layout is critical to achieve reliable, stable, and efficient performance. A ground plane is required to control EMI and minimize the inductance in power, signal, and return paths.

To minimize EMI and output noise, follow these layout recommendations to ensure proper operation:

# 6.1 General

- Most of the heat removed from the IC is due to the exposed pad (EP) on the bottom of the IC conducting heat into the internal ground planes and the ground plane on the bottom side of the board. Use at least 16 vias for the EP to ground plane connection.
- Do not connect the PGND and AGND traces together on the top layer. The single point connection is made on the layer 2 ground plane.
- Do not put a via directly in front of a high current pin, SW, PGND, or PVIN. This will increase the trace resistance and parasitic inductance.
- Do not place a via in between the input and output capacitor ground connection. Put it to the inside of the output capacitor and in the way of the high di/dt current path.
- Route all power traces on the top layer.
- Place the input capacitors first and put them as close as possible to the IC.

# 6.2 IC

- The 2.2  $\mu$ F ceramic capacitor, which is connected to the AVIN pin, must be located right at the IC. The AVIN pin is very noise sensitive and placement of the capacitor is very critical. Use wide traces to connect to the AVIN and AGND pins.
- The analog ground pin (AGND) must be connected directly to the ground planes. Do not route the SGND pin to the PGND Pad on the top layer.
- Use wide traces to route the input and output power lines.
- Use Layer 5 as an input voltage power plane.
- Layer 2 and the bottom layer (Layer 6) are ground planes.

# 6.3 Input Capacitor

- A 10 µF X5R or X7R dielectrics ceramic capacitor is recommended on each of the PVIN pins for bypassing.
- Place the input capacitors on the same side of the board and as close to the IC as possible.
- Keep both the PVIN pin and PGND connections short.

- If possible, place vias to the ground plane close to the each input capacitor ground terminal, but not in the way of the high di/dit current path.
- Use either X7R or X5R dielectric input capacitors. Do not use Y5V or Z5U type capacitors.
- Do not replace the ceramic input capacitor with any other type of capacitor. Any type of capacitor can be placed in parallel with the input capacitor.
- In "Hot-Plug" applications, a Tantalum or Electrolytic bypass capacitor must be used to limit the over-voltage spike seen on the input supply with power is suddenly applied.

# 6.4 Inductor

- Keep the inductor connection to the switch node (SW) short.
- Do not route any digital lines underneath or close to the inductor.
- To minimize noise, place a ground plane underneath the inductor.

# 6.5 Output Capacitor

- Use a wide trace to connect the output capacitor ground terminal to the input capacitor ground terminal.
- The OUT[1-6] trace should be separate from the power trace and connected as close as possible to the output capacitor. Sensing a long high-current load trace can degrade the DC load regulation.

# 6.6 **Proper Termination of Unused Pins**

Many designs will not require all six DC/DC output voltages. In these cases, the unused pin must be connected to either VIN or GND. The schematic in Figure 6-1 shows where to tie the unused pins and Table 6-1 summarizes the connections.

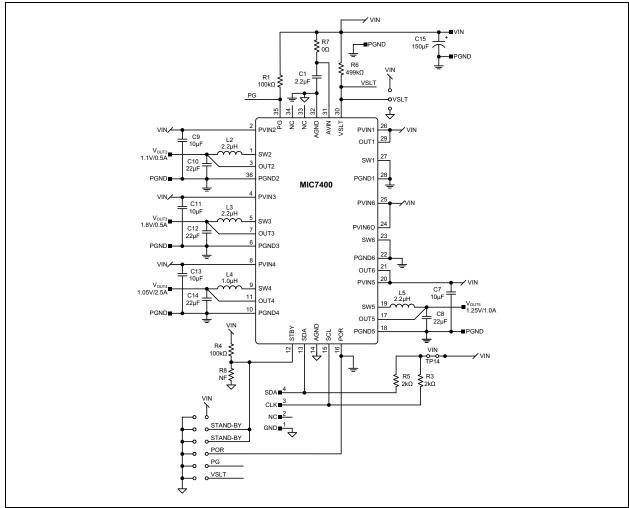


FIGURE 6-1: Connections for Unused Pins.

# TABLE 6-1: SUMMARIZATION OF UNUSED PIN CONNECTIONS

Unused	VIN	PGND
Boost	PVIN6, PGIN6O, VOUT6	PGND6, SW6
Buck	PVIN[x], VOUT[x}	PGND[6], SW[x]
POR	_	POR

# 7.0 I<sup>2</sup>C CONTROL REGISTER

The MIC7400 I<sup>2</sup>C Read/Write registers are detailed here. During normal operation, the configuration data can be saved into non-volatile registers in EEPROM by addressing the chip and writing to SAVECONFIG key = 66'h. Saving CONFIG data to EEPROM takes time so the external host should poll the MIC7400 and read the CONFIG bit[1] of EEPROM Ready register 01'h to determine the end of programming.

All transactions start with a control byte sent from the I<sup>2</sup>C master device. The control byte begins with a START condition, followed by a 7-bit slave address. The slave address is seven bits long followed by an eighth bit which is a data direction bit (R/W), a "0" indicates a transmission (WRITE) and a "1" indicates a request for data (READ). A data transfer is always terminated by a STOP condition that is generated by the master.

# 7.1 Serial Port Operation

# 7.1.1 EXTERNAL HOST INTERFACE

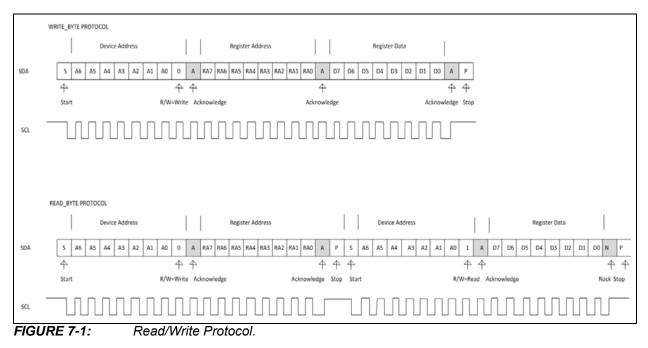
Bidirectional I<sup>2</sup>C port capable of Standard (up to 100 kbits/s), Fast (up to 400 kbits/s), Fast Plus (up to 1 Mbit/s) and High Speed (up to 3.4 Mbit/s) as defined in the I<sup>2</sup>C-Bus Specification.

The MIC7400 acts as an  $I^2C$  slave when addressed by the external host. The MIC7400 slave address uses a fixed 7-bit code and is followed by an R/W bit which is part of the control word that is right after the start bit as shown in Figure 7-1 in the Device Address column.

The MIC7400 can receive multiple data bytes after a single address byte and automatically increments its register pointer to block fill internal volatile memory. Byte data is latched after individual bytes are received so multi-byte transfers could be corrupted if interrupted mid-stream.

No system clock is required by the digital core for I<sup>2</sup>C access from the external host (only the host SCL clock is assumed).

In order to prevent spurious operation of the  $I^2C$ , if a start bit is seen, then any partial communication is aborted and new  $I^2C$  data is allowed. Start bit is when SDA goes low when SCL is high. Stop bit is when SDA goes high when SCL is high. Normal  $I^2C$  exchange is shown in Figure 7-1.



# 7.1.2 SPECIAL HOST I<sup>2</sup>C COMMANDS

The following commands are all 2 byte communications:

- Byte1 = Device address with write bit set, LSB = 0.
- Byte2 = Special key.

Special keys include the following:

- SAVECONFIG Key = 66'h. Saves the shadow register configuration data into EEPROM registers 03'h through 23'h.
- RESET Key = 6A'h. Reloads only NORMAL mode voltage and current limit settings then enables the regulator to NORMAL mode with no soft-start, no sequencing, and no delays. Then it clears the STANDBY register bit 6 in register 03'h.
- RELOAD Key = 6B'h. Reloads all data from EEPROM into the shadow registers. No other actions are performed, including soft-start, sequencing, and delay.
- REBOOT Key = 6C'h. Turns all regulators OFF, reloads EEPROM data into shadow registers, then re-sequences the regulators with the programmed soft-start and sequence delays.
- SEQUENCE Key = 6D'h. Turns all regulators OFF, restarts the sequencer including soft-start and sequence delays.

OBS: In order to use the Special Keys, FORCE\_CLK\_ON bit 2 of the Internal Clock control register (0x2F'h) must be set to "1". After the action has completed, the FORCE\_CLK\_ON bit can be cleared by writing "0".

# 8.0 REGISTER SETTINGS DESCRIPTIONS

#### 8.1 Power Good Register (00'h)

This register indicates when the regulators 1 - 6 output voltage is above 91% of the target value. The MIC7400 deglitches the input signal for 50 µs in order to prevent false events. The global PG pin indicator is functional 'AND' of all the power good indicators during sequencing. Once the power-up sequence is complete, the global power good output stays high unless an output falls below its power-good limit, a thermal fault occurs, the input voltage drops below the lower UVLO threshold or an output is turned OFF by setting the enable register bit to "0" if the PGOOD\_MASK[x] bit is set to "0".

Register Name	PGOOD1-6_REG			Power Good Status Register		
Address		—		0x0	00'h	
Field	Bit	R/W	Default	Descr	iption	
PGOOD1	0	R	0	Power Good indica	ator for Regulator 1	
PGOODT	0	к	0	0 = Buck Not Valid	1 = Buck Valid	
PGOOD2	1	R	0	Power Good indica	ator for Regulator 2	
PGOODZ	I	ĸ	0	0 = Buck Not Valid	1 = Buck Valid	
PGOOD3	2	R	0	Power Good indicator for Regulator 3		
FGOOD3	Z	ĸ	0	0 = Buck Not Valid	1 = Buck Valid	
PGOOD4	3	R	0	Power Good indicator for Regulator 4		
FGOOD4	3	ĸ	0	0 = Buck Not Valid	1 = Buck Valid	
PGOOD5	4			Power Good indica	ator for Regulator 5	
PGOODS	4	R	0	0 = Buck Not Valid	1 = Buck Valid	
PGOOD6	5	R	0	Power Good indica	ator for Regulator 6	
FGOOD6	5	ĸ	0	0 = Boost Not Valid	1 = Boost Valid	
Reserved	6	R/W	0	Not Used	_	
Reserved	7	R/W	0	Not Used	_	

<b>TABLE 8-1</b> :	POWER GOOD STATUS REGISTER
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# 8.2 EEPROM-Ready Register (01'h)

This register indicates the status of EEPROM to external I<sup>2</sup>C host.

The READY bit = 1 when the Trim and Configuration data have been loaded into core from EEPROM after reset, reboot or reload and the chip is ready for operation. If the SAVE1 bit in register 04'h is read in as logic 1, the configuration registers will not be loaded from the EEPROM memory and the READY bit will still get set indicating that any startup procedure involving the EEPROM memory is complete. The READY bit will be set to 1 after loading or attempting to load Trim and Configuration data from EEPROM into volatile memory. The Trim data will always be loaded and if SAVE1 bit in register 04'h is set to logic 0, Configuration data is also loaded. Regardless of the SAVE1 bit being set or not, after the loading operation the READY bit is set to 1.

The CONFIG bit = 1 when the Configuration data have been saved to EEPROM after the SAVECONFIG Code is issued from the Host. If CONFIG=1 before the SAVECONFIG code is issued, CONFIG will be cleared immediately and then will be set to logic 1 again once all Configuration data is written to the EEPROM memory.

The EEPREAD and EEPWRITE bits indicate if an EEPROM read or write fault has occurred. These bits should be read and cleared prior to reloading data from the EEPROM memory.

Register Name	STATUS_REG			EEPROM Sta	atus Register	
Address				0x0	1'h	
Field	Bit	R/W	Default	Descr	iption	
READY	0	R	0	Indicate ready for operation when the trim and configuration dat has been loaded.		
				0 = Data not loaded	1 = Chip ready	
	1			Indicate Configuration saved to EEPROM		
CONFIG	I	R	0	0 = Configuration not saved	1 = Configuration saved	
Reserved	2	R	0	Not Used		
Reserved	3	R/W	0	Not Used		
Reserved	4	R/W	0	Not Used	_	
Reserved	5	R/W	0	Not Used		
	0	R/W	0	EEPRO	M Read	
EEPREAD	6	K/VV	0	0 = No Fault	1 = Fault	
	7	7 R/W	0	EEPRO	M Write	
EEPWRITE	1		0	0 = No Fault	1 = Fault	

#### TABLE 8-2: EEPROM STATUS REGISTER

#### 8.3 Fault Registers (02'h)

This register indicates the overcurrent flag for each regulator and one global overtemperature (OT). These register bits are set by an overcurrent condition and reset by writing a logic "0" to each bit by the  $I^2C$  host. The respective channel must be restarted to enter normal functionality in order to successfully clear the over current fault.

If the fault condition persists, the bit will be set to logic "1" again immediately by the MIC7400 after it is written to logic "0" by the host.

TABLE 8-3:OVERCURRENT STATUS FAULT REGISTER

Register Name	F	AULT_RE	G	Overcurrent Status Fault Register		
Address	_			0x0	2'h	
Field	Bit	R/W	Default	Descr	iption	
REG10C	0	R/W	0	Regulator 1	Overcurrent	
REGIOC	0	FK/ VV	0	0 = No Fault	1 = Fault	
REG2OC	1		0	Regulator 2	Overcurrent	
REG2UC	I	R/W	0	0 = No Fault	1 = Fault	
REG3OC	2	2 R/W 0		Regulator 3	Overcurrent	
REGSOC	2	FK/ VV	0	0 = No Fault	1 = Fault	
	3	R/W	0	Regulator 4 Overcurrent		
REG4OC	3	FK/ VV	0	0 = No Fault	1 = Fault	
DECEOC	4	R/W	0	Regulator 5	Overcurrent	
REG5OC	4	R/W	0	0 = No Fault	1 = Fault	
REG6OC	5	R/W	0	Regulator 6	Overcurrent	
REGOUC	5	FK/ VV	0	0 = No Fault	1 = Fault	
Reserved	6	R/W	0	Reserved		
ОТ	7	R/W	0	Overtem	perature	
ОТ	/	rt/VV	0	0 = No Fault	1 = Fault	

#### 8.4 Standby Register (03'h)

This register controls standby mode operation. Global standby mode can either be enabled by  $I^2C$  or by changing the logic state of the STBY input pin. Global standby is controlled by the STBY\_MODEB bit. When STBY\_MODEB [6] = 1 then the regulators output voltages are set to their normal mode output voltage settings, (05'h - 0A'h) registers. When STBY\_MODEB [6] = 0 then regulators output voltages are set to the standby mode output voltage settings, (0B'h - 10'h) registers. If STBY [1-6] register is set to logic "0", then the output is shut off in standby mode.

The global power good flag is asserted when an output is disabled unless the power good mask bit (PGOOD\_MASK[x]) is set to 1.

Register Name	STBY_CTRL_REG			Standby	Register
Address	_			0x0	)3'h
Field	Bit	R/W	Default	Descr	ription
	0		4	Regulator 1 Stand	by Voltage Control
STBY1	0	R/W	1	0 = OFF	1 = ON
STBY2	1	R/W	1	Regulator 2 Stand	by Voltage Control
31012	I	FC/ V V	1	0 = OFF	1 = ON
STBY3	2	R/W	1	Regulator 3 Stand	by Voltage Control
31013	Z	FC/ V V		0 = OFF	1 = ON
STBY4	3	R/W	1	Regulator 4 Stand	by Voltage Control
31014	3	FC/ V V	1	0 = OFF	1 = ON
STBY5	4	R/W	1	Regulator 5 Stand	by Voltage Control
31615	4	FX/ V V	1	0 = OFF	1 = ON
STBY6	5	R/W	1	Regulator 6 Stand	by Voltage Control
31610	5	FC/ V V	I	0 = OFF	1 = ON
				Global Star	ndby Control
STBY_MODEB	STBY_MODEB 6 R/W		1	0 = All regulators in Standby Mode	1 = All regulators in Normal Mode
				Save Cor	nfiguration
SAVE1	7 R/W		0	0 = Configuration Saved to EEPROM	1 = Configuration Not Saved to EEPROM

TABLE 8-4: STANDBY REGISTER

#### 8.5 Enable/Disable Register (04'h)

This register controls the enable/disable of each DC/DC regulators. When EN(n) bit transitions from "0" to "1", then the regulator(n) is enabled with soft-start unless the STBY\_MODEB register bit in register 03'h is set to logic "0".

The configuration save bit "SAVE1" should be cleared by customer before saving configuration data to EEPROM. This bit is used during power up to indicate via the Status register (00'h) that configuration data has previously been stored.

TABLE 8-5:	ENABLE REGISTER

Register Name	EN_REG			Enable	Enable Register	
Address		—		0x0	)4'h	
Field	Bit	R/W	Default	Description		
	0			Regulator 1 ON/OFF Control bit		
EN1	0	R/W		0 = OFF	1 = ON	
EN2	1	R/W	1	Regulator 2 ON	/OFF Control bit	
EINZ	I	FK/ VV	1	0 = OFF	1 = ON	
EN2	2	R/W	1	Regulator 3 O	N/OFF Control	
EN3	2	FK/ VV		0 = OFF	1 = ON	

Register Name	EN_REG			Enable	Enable Register	
Address	_			0x0	)4'h	
Field	Bit	R/W	Default	Descr	iption	
EN4	3	R/W	Regulator 4 ON/OFF Control		N/OFF Control	
EIN4	3	R/W	1	0 = OFF	1 = ON	
EN5	4	R/W	1	Regulator 5 ON/OFF Control		
ENS	4	r///	1	0 = OFF	1 = ON	
EN6	5	R/W	1	Regulator 6 O	N/OFF Control	
EINO	5	r///		0 = OFF	1 = ON	
Reserved	6	R/W	0	Not Used	_	
Reserved	7	R/W	0	Not Used	_	

#### TABLE 8-5: ENABLE REGISTER (CONTINUED)

# 8.6 Regulator Output Voltage Setting NORMAL Mode (05'h – 09'h)

One register for each regulator output (OUT1 – OUT5). Sets output voltage of regulator for NORMAL mode operation.

Register Name		OUT1	-5_REG		DVC Register	s for OUT[1-5]	
Address			_	OUT1 =	0x05'h; OUT2 = OUT4 = 0x08'h;	0x06'h; OUT3 = OUT5 = 0x09'h	= 0x07'h
Field	Bit	R/W	Default		Descr	iption	
					Output Voltage se 'C from 3.3V to 0		
				000000 = 3.30V	010000 = 2.50V	100000 = 1.70V	110000 = 0.90V
				000001 = 3.25V	010001 = 2.45V	100001 = 1.65V	110001 = 0.85V
				000010 = 3.20V	010010 = 2.40V	100010 = 1.60V	110010 = 0.80V
			OUT1 = 011110	000011 = 3.15V	010011 = 2.35V	100011 = 1.55V	110011 = 0.80V
			(1.8V) OUT2 = 101100 (1.1V)	000100 = 3.10V	010100 = 2.30V	100100 = 1.50V	110100 = 0.80V
				000101 = 3.05V	010101 = 2.25V	100101 = 1.45V	110101 = 0.80V
	5:0	R/W	OUT3 = 011110	000110 = 3.00V	010110 = 2.20V	100110 = 1.40V	110110 = 0.80V
OUT[1-5]	5.0	R/W	(1.8V)	000111 = 2.95V	010111 = 2.15V	100111 = 1.35V	110111 = 0.80V
		OUT4 = 1011	OUT4 = 101101	001000 = 2.90V	011000 = 2.10V	101000 = 1.30V	111000 = 0.80V
			(1.05V) OUT5 = 101001	001001 = 2.85V	011001 = 2.05V	101001 = 1.25V	111001 = 0.80V
			(1.25V)	001010 = 2.80V	011010 = 2.00V	101010 = 1.20V	111010 = 0.80V
			(1.201)	001011 = 2.75V	011011 = 1.95V	101011 = 1.15V	111011 = 0.80V
				001100 = 2.70V	011100 = 1.90V	101100 = 1.10V	111100 = 0.80V
				001101 = 2.65V	011101 = 1.85V	101101 = 1.05V	111101 = 0.80V
				001110 = 2.60V	011110 = 1.80V	101110 = 1.00V	111110 = 0.80V
				001111 = 2.55V	011111 = 1.75V	101111 = 0.95V	111111 = 0.80V
—	6	—	0	Not	Used	_	_
—	7	_	0	Not	Used	-	_

#### TABLE 8-6:DVC REGISTERS FOR OUT[1 - 5]

# 8.7 Boost Regulator Output Voltage Setting NORMAL Mode (0A'h)

Sets output voltage of the boost regulator (OUT6) in NORMAL mode operation.

Register Name	OUT6_REG				DVC Registers			
Address		—			0x0	A'h		
Field	Bit	R/W	Default		Descr	iption		
				DVC	from 14V to 7V i	n 200 mV decren	nents	
				000000 = 14.0V	010000 = 10.8V	100000 = 7.6V	110000 = 7.0V	
				000001 = 13.8V	010001 = 10.6V	100001 = 7.4V	110001 = 7.0V	
				000010 = 13.6V	010010 = 10.4V	100010 = 7.2V	110010 = 7.0V	
				000011 = 13.4V	010011 = 10.2V	100011 = 7.0V	110011 = 7.0V	
			001010 (12V)	000100 = 13.2V	010100 = 10.0V	100100 = 7.0V	110100 = 7.0V	
		R/W		000101 = 13.0V	010101 = 9.8V	100101 = 7.0V	110101 = 7.0V	
				000110 = 12.8V	010110 = 9.6V	100110 = 7.0V	110110 = 7.0V	
OUT6	5:0			000111 = 12.6V	010111 = 9.4V	100111 = 7.0V	110111 = 7.0V	
				001000 = 12.4V	011000 = 9.2V	101000 = 7.0V	111000 = 7.0V	
				001001 = 12.2V	011001 = 9.0V	101001 = 7.0V	111001 = 7.0V	
				001010 = 12.0V	011010 = 8.8V	101010 = 7.0V	111010 = 7.0V	
				001011 = 11.8V	011011 = 8.6V	101011 = 7.0V	111011 = 7.0V	
				001100 = 11.6V	011100 = 8.4V	101100 = 7.0V	111100 = 7.0V	
				001101 = 11.4V	011101 = 8.2V	101101 = 7.0V	111101 = 7.0V	
				001110 = 11.2V	011110 = 8.0V	101110 = 7.0V	111110 = 7.0V	
				001111 = 11.0V	011111 = 7.8V	101111 = 7.0V	111111 = 7.0V	
—	6	_	0	Not	Used	_	_	
—	7	—	0	Not	Used	-	_	

#### TABLE 8-7: DVC REGISTERS FOR OUT6

# 8.8 Regulator Voltage Setting STBY Mode (0B'h – 0F'h)

This register is used to sets the output voltage of regulators 1 - 5 in STBY mode operation.

#### TABLE 8-8: STANDBY REGISTERS

Register Name		STBY_O	UT1-5_REG	Standby DVC Registers			
Address			_	OUT1 =	,	0x0C'h; OUT3 = OUT5 = 0x0F'h	= 0x0D'h
Field	Bit	R/W	Default		Descr	iption	
						etting of OUT[1-5 .8V in –50 mV ste	
				000000 = 3.30V	010000 = 2.50V	100000 = 1.70V	110000 = 0.90V
				000001 = 3.25V	010001 = 2.45V	100001 = 1.65V	110001 = 0.85V
				000010 = 3.20V	010010 = 2.40V	100010 = 1.60V	110010 = 0.80V
			OUT1 = 011110	000011 = 3.15V	010011 = 2.35V	100011 = 1.55V	110011 = 0.80V
			(1.8V) OUT2 = 101100 (1.1V)	000100 = 3.10V	010100 = 2.30V	100100 = 1.50V	110100 = 0.80V
				000101 = 3.05V	010101 = 2.25V	100101 = 1.45V	110101 = 0.80V
SB_OUT	5:0	R/W	OUT3 = 011110	000110 = 3.00V	010110 = 2.20V	100110 = 1.40V	110110 = 0.80V
[1-5]	5.0	F\/ VV	(1.8V)	000111 = 2.95V	010111 = 2.15V	100111 = 1.35V	110111 = 0.80V
			OUT4 = 101101	001000 = 2.90V	011000 = 2.10V	101000 = 1.30V	111000 = 0.80V
			(1.05V) OUT5 = 101001	001001 = 2.85V	011001 = 2.05V	101001 = 1.25V	111001 = 0.80V
			(1.25V)	001010 = 2.80V	011010 = 2.00V	101010 = 1.20V	111010 = 0.80V
			(1.201)	001011 = 2.75V	011011 = 1.95V	101011 = 1.15V	111011 = 0.80V
				001100 = 2.70V	011100 = 1.90V	101100 = 1.10V	111100 = 0.80V
				001101 = 2.65V	011101 = 1.85V	101101 = 1.05V	111101 = 0.80V
				001110 = 2.60V	011110 = 1.80V	101110 = 1.00V	111110 = 0.80V
				001111 = 2.55V	011111 = 1.75V	101111 = 0.95V	111111 = 0.80V
_	6	_	0	Not I	Jsed	-	_
	7	_	0	Not I	Jsed		_

#### 8.9 Boost Regulator Output Voltage Setting STBY Mode (10'h)

Sets output voltage of the boost regulator (OUT6) for STBY mode operation.

Register Name	STBY_OUT6_REG				DVC Registers			
Address		_			0x1	0'h		
Field	Bit	R/W	Default		Descr	iption		
				DVC	from 14V to 7V i	n 200 mV decren	nents	
				000000 = 14.0V	010000 = 10.8V	100000 = 7.6V	110000 = 7.0V	
				000001 = 13.8V	010001 = 10.6V	100001 = 7.4V	110001 = 7.0V	
				000010 = 13.6V	010010 = 10.4V	100010 = 7.2V	110010 = 7.0V	
				000011 = 13.4V	010011 = 10.2V	100011 = 7.0V	110011 = 7.0V	
	5:0	R/W	001010 (12V)	000100 = 13.2V	010100 = 10.0V	100100 = 7.0V	110100 = 7.0V	
				000101 = 13.0V	010101 = 9.8V	100101 = 7.0V	110101 = 7.0V	
				000110 = 12.8V	010110 = 9.6V	100110 = 7.0V	110110 = 7.0V	
SB_OUT6				000111 = 12.6V	010111 = 9.4V	100111 = 7.0V	110111 = 7.0V	
				001000 = 12.4V	011000 = 9.2V	101000 = 7.0V	111000 = 7.0V	
				001001 = 12.2V	011001 = 9.0V	101001 = 7.0V	111001 = 7.0V	
				001010 = 12.0V	011010 = 8.8V	101010 = 7.0V	111010 = 7.0V	
				001011 = 11.8V	011011 = 8.6V	101011 = 7.0V	111011 = 7.0V	
				001100 = 11.6V	011100 = 8.4V	101100 = 7.0V	111100 = 7.0V	
				001101 = 11.4V	011101 = 8.2V	101101 = 7.0V	111101 = 7.0V	
				001110 = 11.2V	011110 = 8.0V	101110 = 7.0V	111110 = 7.0V	
				001111 = 11.0V	011111 = 7.8V	101111 = 7.0V	111111 = 7.0V	
—	6	_	0	Not	Used	<u> </u>		
—	7	_	0	Not	Used	_	_	

#### TABLE 8-9:STANDBY DVC REGISTER FOR OUT6

#### 8.10 Sequence Register (11'h)

Each regulator can be assigned to start in any one of six sequencing slots (1 to 6). If starting in slot 1, the regulator starts immediately. If starting in any other slot, the regulator must wait for the PGOOD = 1 flags of all regulators assigned to the preceding slot and then wait for the specified delay time (register 17'h) i.e., all PGOODs in preceding state flag then the delay timer is started and when delay completes the regulator is enabled.

Each regulator will delay its startup (after the appropriate preceding PGOOD flags) by the delay set in the Delay Register (17'h), unless the regulator is assigned to sequence state 0.

If all default Enable bits = 0 the IC starts up, but no outputs are enabled.

Sequencing is only used during initial startup, and not used when outputs are enabled via I<sup>2</sup>C command. If outputs are enabled via I<sup>2</sup>C, then soft-start is still active, but start-up delays (timed from preceding PGOODs) are not.

Register Name	SEQ1_REG			Sequence	e Register		
Address	—		ddress			0x1	11'h
Field	Bit R/W Default		R/W Default Description		iption		
REG1SQ1	0	R/W	0	0 = No Start	1 = Regulator 1 will Start in Sequence State 1		
REG2SQ1	1	R/W	0	0 = No Start	1 = Regulator 2 will Start in Sequence State 1		
REG3SQ1	2	R/W	0	0 = No Start	1 = Regulator 3 will Start in Sequence State 1		

TABLE 8-10: SEQUENCE STATE 1 REGISTER

# TABLE 8-10: SEQUENCE STATE 1 REGISTER (CONTINUED)

Register Name	SEQ1_REG		3	Sequence	Sequence Register		
Address		—		0x1	l1'h		
Field	Bit R/W Default		Default	Description			
REG4SQ1	3	R/W	1	0 = No Start	1 = Regulator 4 will Start in Sequence State 1		
REG5SQ1	4	R/W	0	0 = No Start	1 = Regulator 5 will Start in Sequence State 1		
REG6SQ1	5	R/W	0	0 = No Start	1 = Regulator 6 will Start in Sequence State 1		
_	6	R/W	0	Reserved			
—	7	R/W	0	Reserved			

#### TABLE 8-11: SEQUENCE STATE 2 REGISTER

Register Name	SEQ2_REG		3	Sequence	e Register
Address		—		0x1	2'h
Field	Bit	R/W	Default	Descr	iption
REG1SQ2	0	R/W	0	0 = No Start	1 = Regulator 1 will Start in Sequence State 2
REG2SQ2	1	R/W	1	0 = No Start	1 = Regulator 2 will Start in Sequence State 2
REG3SQ2	2	R/W	1	0 = No Start	1 = Regulator 3 will Start in Sequence State 2
REG4SQ2	3	R/W	0	0 = No Start	1 = Regulator 4 will Start in Sequence State 2
REG5SQ2	4	R/W	0	0 = No Start	1 = Regulator 5 will Start in Sequence State 2
REG6SQ2	5	R/W	0	0 = No Start	1 = Regulator 6 will Start in Sequence State 2
	6	R/W	0	Rese	erved
	7	R/W	0	Rese	erved

#### TABLE 8-12: SEQUENCE STATE 3 REGISTER

Register Name	SEQ3_REG		ne SEQ3_REG Sequence Register		e Register
Address		—		0x13'h	
Field	Bit	R/W	Default	Desci	iption
REG1SQ3	0	R/W	1	0 = No Start	1 = Regulator 1 will Start in Sequence State 3
REG2SQ3	1	R/W	0	0 = No Start	1 = Regulator 2 will Start in Sequence State 3
REG3SQ3	2	R/W	0	0 = No Start	1 = Regulator 3 will Start in Sequence State 3
REG4SQ3	3	R/W	0	0 = No Start	1 = Regulator 4 will Start in Sequence State 3
REG5SQ3	4	R/W	0	0 = No Start	1 = Regulator 5 will Start in Sequence State 3
REG6SQ3	5	R/W	0	0 = No Start	1 = Regulator 6 will Start in Sequence State 3

# TABLE 8-12: SEQUENCE STATE 3 REGISTER (CONTINUED)

Register Name	SEQ3_REG			Sequence Register
Address	—			0x13'h
Field	Bit R/W Default		Default	Description
	6	R/W	0	Reserved
	7	R/W	0	Reserved

#### TABLE 8-13: SEQUENCE STATE 4 REGISTER

Register Name	SEQ4_REG		3	Sequence	e Register
Address		_		0x14'h	
Field	Bit	R/W	Default	Desci	ription
REG1SQ4	0	R/W	0	0 = No Start	1 = Regulator 1 will Start in Sequence State 4
REG2SQ4	1	R/W	0	0 = No Start	1 = Regulator 2 will Start in Sequence State 4
REG3SQ4	2	R/W	0	0 = No Start	1 = Regulator 3 will Start in Sequence State 4
REG4SQ4	3	R/W	0	0 = No Start	1 = Regulator 4 will Start in Sequence State 4
REG5SQ4	4	R/W	1	0 = No Start	1 = Regulator 5 will Start in Sequence State 4
REG6SQ4	5	R/W	0	0 = No Start	1 = Regulator 6 will Start in Sequence State 4
	6	R/W	0	Reserved	
—	7	R/W	0	Reserved	

#### TABLE 8-14: SEQUENCE STATE 5 REGISTER

Register Name	SEQ5_REG			Sequence	e Register
Address		_		0x15'h	
Field	Bit	R/W	Default	Desci	ription
REG1SQ5	0	R/W	0	0 = No Start	1 = Regulator 1 will Start in Sequence State 5
REG2SQ5	1	R/W	0	0 = No Start	1 = Regulator 2 will Start in Sequence State 5
REG3SQ5	2	R/W	0	0 = No Start	1 = Regulator 3 will Start in Sequence State 5
REG4SQ5	3	R/W	0	0 = No Start	1 = Regulator 4 will Start in Sequence State 5
REG5SQ5	4	R/W	0	0 = No Start	1 = Regulator 5 will Start in Sequence State 5
REG6SQ5	5	R/W	0	0 = No Start	1 = Regulator 6 will Start in Sequence State 5
—	6	R/W	0	Reserved	
—	7	R/W	0	Rese	erved

Register Name	SEQ6_REG			Sequence	e Register
Address		_	0x16'h		l6'h
Field	Bit	R/W	Default	Descr	ription
REG1SQ6	0	R/W	0	0 = No Start	1 = Regulator 1 will Start in Sequence State 6
REG2SQ6	1	R/W	0	0 = No Start	1 = Regulator 2 will Start in Sequence State 6
REG3SQ6	2	R/W	0	0 = No Start	1 = Regulator 3 will Start in Sequence State 6
REG4SQ6	3	R/W	0	0 = No Start	1 = Regulator 4 will Start in Sequence State 6
REG5SQ6	4	R/W	0	0 = No Start	1 = Regulator 5 will Start in Sequence State 6
REG6SQ6	5	R/W	1	0 = No Start	1 = Regulator 6 will Start in Sequence State 6
—	6	R/W	0	Reserved	
—	7	R/W	0	Reserved	

#### TABLE 8-15:SEQUENCE STATE 6 REGISTER

#### 8.11 Delay Register (17'h)

The STDEL register sets the delay between powering up of each regulator at initial power up (see Figure 5-1). Once all the internal power good registers PGOOD[1-6] are all "1", then the global PG pin goes high without delay.

The PORDEL register sets the delay for the POR flag pin. The POR delay time starts as soon as AVIN pin voltage rises above the system UVLO upper threshold set by the PORUP register (21'h). The POR output goes low without delay if AVIN falls below the lower UVLO threshold set by the PORDN register (22'h).

Register Name	DELAY_CNTL_REG			Delay Register			
Address		—			0x17'h		
Field	Bit	R/W	Default		Descr	ription	
			004	Delay T	ime from 0 ms to	o 7 ms in 1 ms ind	crement
STDEL	2:0	R/W	001 (1 ms)	000 = 0 ms	010 = 2 ms	100 = 4 ms	110 = 6 ms
			(1115)	001 = 1 ms	011 = 3 ms	101 = 5 ms	111 = 7 ms
		R/W	00011 (20 ms)	Delay Tir	me from 5 ms to	160 ms in 5 ms ir	ncrement
				00000 = 5ms	01000 = 45ms	10000 = 85ms	11000 = 125ms
				00001 = 10ms	01001 = 50ms	10001 = 90ms	11001 = 130ms
				00010 = 15ms	01010 = 55ms	10010 = 95ms	11010 = 135ms
PORDEL	7:3			00011 = 20ms	01011 = 60ms	10011 = 100ms	11011 = 140ms
			(201110)	00100 = 25ms	01100 = 65ms	10100 = 105ms	11100 = 145ms
			-	00101 = 30ms	01101 = 70ms	10101 = 110ms	11101 = 150ms
				00110 = 35ms	01110 = 75ms	10110 = 115ms	11110 = 155ms
				00111 = 40ms	01111 = 80ms	10111 = 120ms	11111 = 160ms

#### TABLE 8-16: DELAY REGISTER

# 8.12 Soft-Start Registers (18'h – 1A'h)

When regulator(n) is turned on from either the Enable Register (04'h) in NORMAL mode or from the Standby Register (03'h) in STANDBY mode, the three REG(n)SS soft-start bits are used to control both the rising and falling ramp rate of the outputs.

In NORMAL mode, the outputs are stepped from the current regulator voltage settings to a newly programmed regulator voltage setting or to the default value.

On power-up, the regulator voltage output is set to the lowest possible voltage setting, which is 3F'h. The voltage regulator will change by one step or increment at a time. The amount of time between each step is controlled by the soft-start registers. Table 8-17 details the amount of time for each encoded soft-start value.

TABLE 8-17: SOFT-START REGISTER SPEED SETTINGS							
Register	R/W	Default	Description				
	SS_SPEED = 0 R/W		Soft-Start Time from 4 µs to 512 µs				
SS_SPEED = 0		000	000 = 4 µs	010 = 16 µs	100 = 64 µs	110 = 256 µs	
			001 = 8 µs	011 = 32 µs	101 = 128 µs	111 = 512 µs	
		000	S	Soft-Start Time fro	om 8 µs to 1024 µ	ıs	
SS_SPEED = 1	R/W		000 = 8 µs	010 = 32 µs	100 = 128 µs	110 = 512 µs	
			001 = 16 µs	011 = 64 µs	101 = 256 µs	111 = 1024 μs	

 TABLE 8-17:
 SOFT-START REGISTER SPEED SETTINGS

Register Name	SS1-2_REG		3	Soft-Start Register for $V_{OUT1}$ and $V_{OUT2}$
Address	_		— 0x18'h	
Field	Bit R/W Default		Default	Description
REG1SS	2:0	R/W	001 (8 μs)	OUT1 Soft-Start Time. See Table 8-17 for Soft-Start settings.
REG2SS	5:3	R/W	001 (8 μs)	OUT2 Soft-Start Time. See Table 8-17 for Soft-Start settings.
—	6	R/W	0	Reserved
SS_SPEED	7	R/W	0	Sets the speed of the clock to slow or fast for different clock divi- sion, see Table 8-17. 0 = Slow speed; 1 = Fast speed.

#### TABLE 8-19: SOFT-START REGISTER OUT3 AND OUT4

Register Name	SS3-4_REG			Soft-Start Register for $V_{OUT3}$ and $V_{OUT4}$		
Address	—			0x19'h		
Field	Bit	R/W	Default	Description		
REG3SS	2:0	R/W	001 (8 μs)	OUT3 Soft-Start Time. See Table 8-17 for Soft-Start settings.		
REG4SS	5:3	R/W	001 (8 μs)	OUT4 Soft-Start Time. See Table 8-17 for Soft-Start settings.		
—	6	R/W	0	Reserved		
—	7	R/W	0	Reserved		

Register Name	5	685-6_REC	6	Soft-Start Register for $V_{OUT5}$ and $V_{OUT6}$		
Address	—			0x1A'h		
Field	Bit	R/W	Default	Description		
REG5SS	2:0	R/W	001 (8 μs)	OUT5 Soft-Start Time. See Table 8-17 for Soft-Start settings.		
REG6SS	5:3	R/W	010 (16 μs)	OUT6 Soft-Start Time. See Table 8-17 for Soft-Start settings.		
	6	R/W	0	Reserved		
	7	R/W	0	Reserved		

#### TABLE 8-20: SOFT-START REGISTER OUT5 AND OUT6

# 8.13 Current-Limit (Normal Mode) Registers (1B'h – 1D'h)

This register is used to set the current limit for each DC/DC regulator in normal mode operation.

Register Name	ILI	MIT_1-2_R	EG	Curre	nt-Limit Registe	r for V <sub>OUT1</sub> and	V <sub>OUT2</sub>
Address	— 0x1B'h						
Field	Bit	R/W	Default		Descr	iption	
REG1CL 3:0				Normal current-	•	<sup>.</sup> 1 from 8.6A to 1 ents	.1A in 0.5A dec-
	5.44	1001	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A	
	3:0	R/W	(4.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A
				Normal current-	•	<sup>-</sup> 2 from 8.6A to 1 ents	.1A in 0.5A dec-
DEODOI		DAA	1001	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A
REG2CL	7:4	R/W	(4.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A

# TABLE 8-21: CURRENT-LIMIT REGISTER IOUT1 AND IOUT2

# TABLE 8-22: CURRENT-LIMIT REGISTER IOUT3 AND IOUT4

Register Name	ILI	MIT_3-4_R	EG	Current-Limit Register for $V_{\mbox{OUT3}}$ and $V_{\mbox{OUT4}}$				
Address	— 0x1C'h							
Field	Bit	R/W	Default	Description				
			1001 (4.1A)	Normal current-	-	<sup>·</sup> 3 from 8.6A to 1 ents	.1A in 0.5A dec-	
REG3CL 3:0				0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A	
	3:0 R	R/W		0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A	
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A	
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A	
				Normal current-	limit for regulator	4 from 8.6A to 1	.1A in 0.5A dec-	
				rements				
	- 4	DAA	0101	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A	
REG4CL	7:4	R/W	(6.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A	
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A	
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A	

Register Name	ILIMIT_5-6_REG			Current-Limit Register for $V_{OUT5}$ and $V_{OUT6}$				
Address		—		0x1D'h				
Field	Bit	R/W	Default		Descr	ription		
				Normal current-limit for regulator 5 from 8.6A to 1.1A in 0.5A dec- rements				
	-	1001	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A		
REG5CL	3:0	R/W	(4.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A	
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A	
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A	
				Current-li	mit from 2.6A to	1.78A in 0.12A de	ecrements	
REG6CL	6:4	R/W	011 (2.24A)	000 = 2.6A	010 = 2.36A	100 = 2.12A	110 = 1.88A	
		(2.247)	001 = 2.48A	011 = 2.24A	101 = 2.00A	111 = 1.76A		
—	7	R/W	0	0 = Currer	nt-Limit On	1 = Currer	nt-Limit Off	

TABLE 8-23: CURRENT-LIMIT REGISTER IOUT5 AND IOUT6

# 8.14 Current-Limit (STBY Mode) Registers (1E'h – 20'h)

This register is used to set the current-limit for each DC/DC regulator when in standby (STBY) mode operation.

TABLE 8-24:	STANDBY	CURRENT-LIMIT REGISTER IOUT1	AND IOUT2
-------------	---------	------------------------------	-----------

Register Name	STBY	_ILIMIT_1-	2_REG	Standby C	urrent-Limit Re	gister for V <sub>OUT1</sub>	and V <sub>OUT2</sub>		
Address		— 0x1E'h							
Field	Bit	R/W	Default		Description				
			Standby current	-	r 1 from 8.6A to 1 ents	.1A in 0.5A dec-			
	5.44	1001	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A			
SB1CL	3:0	R/W	(4.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A		
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A		
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A		
				Standby current	-	r 2 from 8.6A to 1 ents	.1A in 0.5A dec-		
00001		5.44	1001	0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A		
SB2CL	7:4	R/W	(4.1A)	0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A		
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A		
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A		

 TABLE 8-25:
 STANDBY CURRENT-LIMIT REGISTER IOUT3 AND IOUT4

Register Name	STBY_ILIMIT_3-4_REG			Standby Current-Limit Register for $V_{OUT3}$ and $V_{OUT4}$			
Address	—			0x1F'h			
Field	Bit	R/W	Default	Description			
		R/W	1001 (4.1A)	Standby current-limit for regulator 3 from 8.6A to 1.1A in 0.5A decrements			
				0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A
SB3CL	3:0			0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A

Register Name	STBY_ILIMIT_3-4_REG			Standby Current-Limit Register for $V_{OUT3}$ and $V_{OUT4}$					
Address	_				0x1	F'h			
Field	Bit	R/W	Default	Description					
			0101 (6.1A)	Standby current-limit for regulator 4 from 8.6A to 1.1A in 0.5A decrements					
		-		0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A		
SB4CL	SB4CL 7:4 R/W	R/W		0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A		
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A		
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A		

# TABLE 8-25: STANDBY CURRENT-LIMIT REGISTER IOUT3 AND IOUT4 (CONTINUED)

#### TABLE 8-26: STANDBY CURRENT-LIMIT REGISTER IOUT5 AND IOUT6

Register Name	STBY_	ILIMIT_5-	6_REG	Standby Current-Limit Register for $V_{OUT5}$ and $V_{OUT6}$				
Address		—		0x20'h				
Field	Bit	R/W	R/W Default Description					
			1001 (4.1A)	Standby current-limit for regulator 5 from 8.6A to 1.1A in 0.5A dec- rements				
00-00	3:0	R/W		0000 = 8.6A	0100 = 6.6A	1000 = 4.6A	1100 = 2.6 A	
SB5CL				0001 = 8.1A	0101 = 6.1A	1001 = 4.1A	1101 = 2.1A	
				0010 = 7.6A	0110 = 5.6A	1010 = 3.6A	1110 = 1.6A	
				0011 = 7.1A	0111 = 5.1A	1011 = 3.1A	1111 = 1.1A	
				Current-limit from 2.6A to 1.78A in 0.12A decrements				
SB6CL	6:4	R/W	011 (2.24A)	000 = 2.6A	010 = 2.36A	100 = 2.12A	110 = 1.88A	
			(2.247)	001 = 2.48A	011 = 2.24A	101 = 2.00A	111 = 1.76A	
—	7	R/W	0	0 = Currer	nt-Limit On	1 = Currer	nt-Limit Off	

#### 8.15 Power-on-Reset (POR) Threshold Voltage Setting Register (21'h and 22'h)

This register is used to set the rising and falling threshold of power-on-reset (POR) comparator. The POR threshold voltage setting is based on the logic level of the VSLT pin in addition to the register bits. Refer to Table 8-16 for POR time delay settings.

Condition	-			Rising and	-	On-Reset Thresh ting	old Voltage
Pin	Bit	R/W	Default	Description			
				:	3.3V to 2.3V in 50	0 mV decrements	6
			00000 = 3.25V	01000 = 2.85V	10000 = 2.45V	11000 = 2.25V	
			00000	00001 = 3.20V	01001 = 2.80V	10001 = 2.40V	11001 = 2.25V
				00010 = 3.15V	01010 = 2.75V	10010 = 2.35V	11010 = 2.25V
VSCLT = 0	4:0	R/W		00011 = 3.10V	01011 = 2.70V	10011 = 2.30V	11011 = 2.25V
				00100 = 3.05V	01100 = 2.65V	10100 = 2.25V	11100 = 2.25V
				00101 = 3.00V	01101 = 2.60V	10101 = 2.25V	11101 = 2.25V
			00110 = 2.95V	01110 = 2.55V	10110 = 2.25V	11110 = 2.25V	
				00111 = 2.90V	01111 = 2.50V	10111 = 2.25V	11111 = 2.25V

#### TABLE 8-27: RISING AND FALLING POWER-ON-RESET THRESHOLD VOLTAGE SETTINGS

Condition	-			Rising and Falling Power-On-Reset Threshold Voltage Setting			
Pin	Bit R/W Default Description						
	4:0	R/W	00000	4.3V to 3.3V in 50 mV decrements			
				00000 = 4.25V	01000 = 3.85V	10000 = 3.45V	11000 = 3.25V
				00001 = 4.20V	01001 = 3.80V	10001 = 3.40V	11001 = 3.25V
				00010 = 4.15V	01010 = 3.75V	10010 = 3.35V	11010 = 3.25V
VSCLT = 1				00011 = 4.10V	01011 = 3.70V	10011 = 3.30V	11011 = 3.25V
				00100 = 4.05V	01100 = 3.65V	10100 = 3.25V	11100 = 3.25V
				00101 = 4.00V	01101 = 3.60V	10101 = 3.25V	11101 = 3.25V
				00110 = 3.95V	01110 = 3.55V	10110 = 3.25V	11110 = 3.25V
				00111 = 3.90V	01111 = 3.50V	10111 = 3.25V	11111 = 3.25V

The three most significant bits [7:5] in registers 21'h and 22'h are used to mask the output voltage power-good flag after the start-up sequenced is finished.

#### TABLE 8-28: POWER-ON-RESET RISING THRESHOLD VOLTAGE SETTING REGISTER (21'H)

Register Name PORUP_REG			G	Power-on-Reset Rising Threshold		
Address	—			0x21'h		
Field	Bit	R/W	Default	Description		
PORUP	4:0	R/W	01011	See Table 8-27		
PGOOD_MASK 1	5	R/W	1	0 = Do not mask PGOOD1	1 = Mask PGOOD1	
PGOOD_MASK 2	6	R/W	1	0 = Do not mask PGOOD2 1 = Mask PGOOD2		
PGOOD_MASK 3	7	R/W	1	0 = Do not mask PGOOD3	1 = Mask PGOOD3	

#### TABLE 8-29: POWER-ON-RESET FALLING THRESHOLD VOLTAGE SETTING REGISTER (22'H)

Register Name	PORDN_REG			Power-on-Reset Falling Threshold		
Address	—			0x22'h		
Field	Bit	R/W	Default Description			
PORDN	4:0	R/W	01101	See Table 8-27		
PGOOD_MASK 4	5	R/W	1	0 = Do not mask PGOOD4	1 = Mask PGOOD4	
PGOOD_MASK 5	6	R/W	1	0 = Do not mask PGOOD5 1 = Mask PGOOD5		
PGOOD_MASK 6	7	R/W	1	0 = Do not mask PGOOD6	1 = Mask PGOOD6	

#### 8.16 Pull-Down When Disabled Register (23'h)

This register is used to set the preference of enabling/disabling a pull-down FET when the DC/DC regulators are disabled. The pull-down value for buck regulators 1 through 5 is  $90\Omega$ . The pull-down current value for the boost regulator 6 is programmable.

Register Name	PULLDN1-6_REG			Pull-Down When Disabled Register		
Address	—			0x23'h		
Field	Bit R/W Default		Default	Description		
PULLD1	0	R/W	0	Enable/Disable the pull-down on Regulator 1 when power down. 0 = No Pull-Down; 1 = Pull-Down		
PULLD2	1	R/W	0	Enable/Disable the pull-down on Regulator 2 when power down. 0 = No Pull-Down; 1 = Pull-Down		
PULLD3	2	R/W	0	Enable/Disable the pull-down on Regulator 3 when power down. 0 = No Pull-Down; 1 = Pull-Down		
PULLD4	3	R/W	0	Enable/Disable the pull-down on Regulator 4 when power down. 0 = No Pull-Down; 1 = Pull-Down		
PULLD5	4	R/W	0	Enable/Disable the pull-down on Regulator 5 when power down. 0 = No Pull-Down; 1 = Pull-Down		
PULLD6C	6:5	R/W	00	Sets Boost Pull-Down Current Level 00 = 148 mA; 01 = 111 mA; 10 = 74 mA; 11 = 37 mA		
PULLD6	7	R/W	0	Enable/Disable the pull-down on Regulator 6 when power down. 0 = No Pull-Down; 1 = Pull-Down		

TABLE 8-30:PULL-DOWN WHEN DISABLED REGISTER

#### 8.17 Internal Clock Control Register

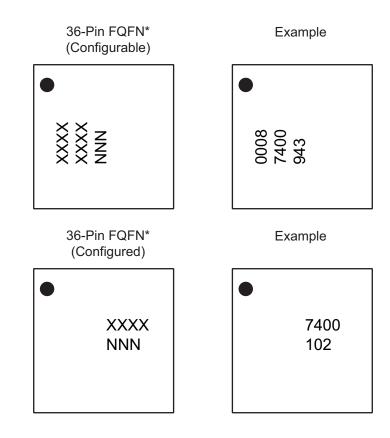
This register houses the Force\_CLK\_ON bit 2 used when sending the special commands keys. Bit 2 of this register is used to set the PMIC clock to permanently be enabled in order to execute the new command. This bit should be cleared after the command has been executed in order to save power (the internal clock logic will shut down the clock automatically when not needed).

Register Name	Force Clock Register		gister	Internal Clock Control Register		
Address	—			0x2F'h		
Field	Bit R/W Default		Default	Description		
—	0	R/W	0	Reserved		
—	1	R/W	0	Reserved		
FORCE CLK_ON	2	R/W	0	0=No action. 1= Force the internal clock to keep running and not be turned off by the power down logic.		
—	3	R/W	0	Reserved		
_	4	R/W	0	Reserved		
—	5	R/W	0	Reserved		
—	6	R/W	0	Reserved		
	7 R/W 0		0	Reserved		

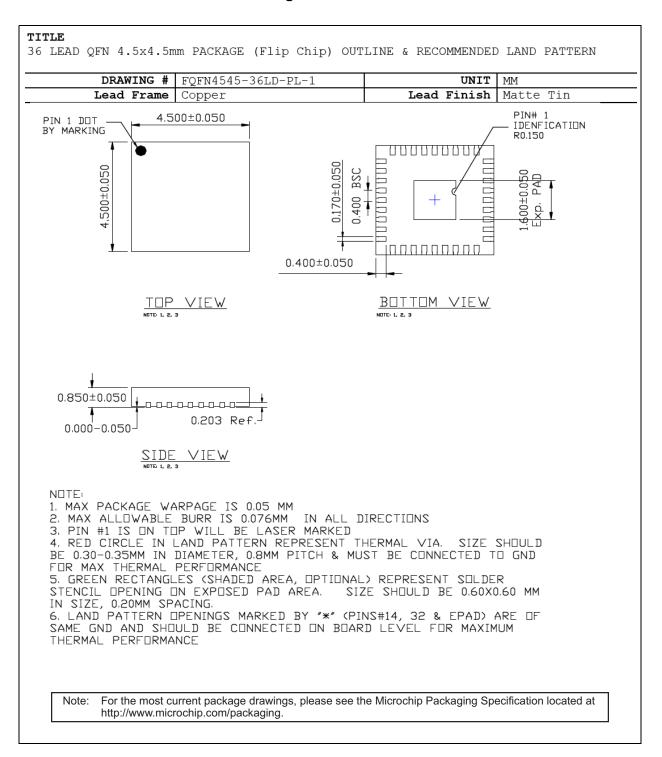
 TABLE 8-31:
 INTERNAL CLOCK CONTROL REGISTER

# 9.0 PACKAGING INFORMATION

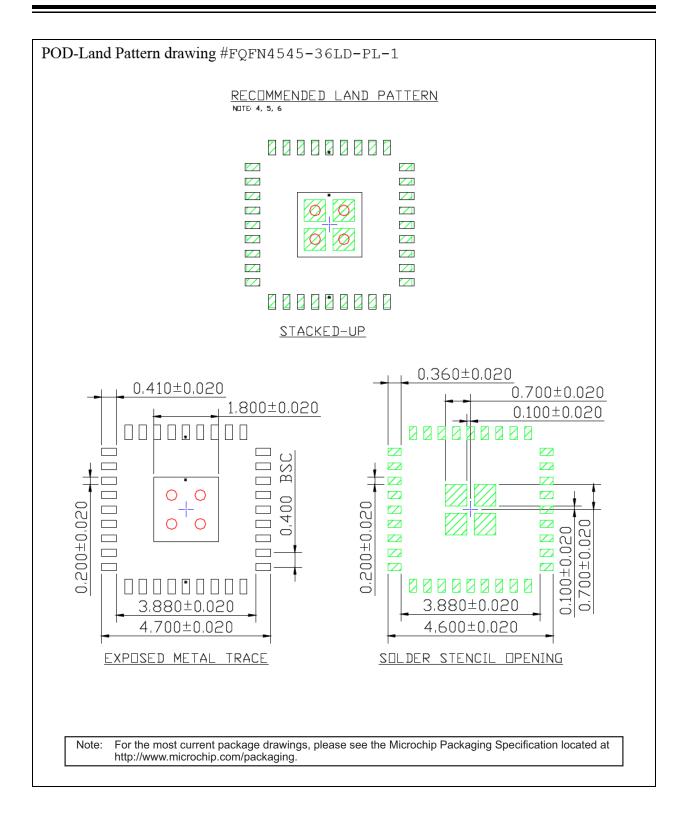
# 9.1 Package Marking Information



Legend	Y YY WW NNN @3 *	Product code or customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC <sup>®</sup> designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package. ' Pin one index is identified by a dot, delta up, or delta down (triangle
	be carried	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available of or customer-specific information. Package may or may not include ate logo.
	Underbar	(_) and/or Overbar (⁻) symbol may not be to scale.



36-Lead 4.5 mm x 4.5 mm FQFN Package Outline and Recommended Land Pattern



# **MIC7400**

NOTES:

# APPENDIX A: REVISION HISTORY

# **Revision A (November 2017)**

- Converted Micrel document MIC7400 to Microchip data sheet DS20005887A.
- Minor text changes throughout.

# **MIC7400**

NOTES:

# **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

	~~~~	v	VV	NV.	Example	es:	
		Regulato	able PMIC, Five	-XX ledia Type	a) MIC7	400YFL-T5:	Configurable PMIC, Five Chan- nel Buck Regulator Plus One Boost with HyperLight Load and I <sup>2</sup> C Control, 1.8V, 1.1V, 1.8V, 1.05V, 1.25V, 12V Output Volt- ages, -40°C to +125°C Temp. Range, 36-Lead FQFN, 500/ Reel
Output Voltages: Junction Temperature Range:	<blank>= XXXX = Y =</blank>		,	V, 12V eting for Options)	b) MIC7 XXXXYF		Configurable PMIC, Five Chan- nel Buck Regulator Plus One Boost with HyperLight Load and I <sup>2</sup> C Control, Configurable Output Voltages, -40°C to +125°C Temp. Range, 36-Lead FQFN, 5,000/Reel
Package: Media Type:	FL = <blank>= T5 = TR =</blank>	36-Lead 4.5 mr 1/Tube 500/Reel 5,000/Reel	n x 4.5 mm FQI	FN	c) MIC74	400YFL:	Configurable PMIC, Five Chan- nel Buck Regulator Plus One Boost with HyperLight Load and I <sup>2</sup> C Control, 1.8V, 1.1V, 1.8V, 1.05V, 1.25V, 12V Output Volt- ages, –40°C to +125°C Temp. Range, 36-Lead FQFN, 1/Tube
					d) MIC7 XXXXYF		Configurable PMIC, Five Chan- nel Buck Regulator Plus One Boost with HyperLight Load and I <sup>2</sup> C Control, Configurable Output Voltages, -40°C to +125°C Temp. Range, 36-Lead FQFN, 500/Reel
					Note 1:	catalog part used for ord the device p	eel identifier only appears in the number description. This identifier is lering purposes and is not printed on package. Check with your Microchip e for package availability with the eel option.

# **MIC7400**

NOTES:

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